## SWITCHING THEORY \& LOGIC DESIGN BIT-11

Contact Hours Lecture : 3, Tutorial : 1, Practical: 2 (online 4 lectures/week) Number of Credits :5
B.Tech (IT)

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## SWITCHING THEORY \& LOGIC DESIGN

| BIT-11 | SWITCHING THEORY \& LOGIC DESIGN |  |  |
| :--- | :--- | :--- | :--- |
| Contact Hours/Week | $:$ | Lecture : 3, Tutorial : 1, Practical: 0 |  |
| UNIT-I |  |  |  |
| Number system, Binary Codes - Weighted and Non-Weighted - Binary Arithmetic Conversion Algorithms - Error <br> Detecting and Error Correcting Codes - Canonical and Standard Boolean Expressions - Truth Tables. | 9 |  |  |
| UNIT-II | K-Map Reduction - Don't Care Conditions - Adders / Subtractors- Carry Look-Ahead Adder - Code Conversion <br> Algorithms - Design of Code Converters - Equivalence Functions. <br> Binary/Decimal Parallel Adder/Subtractorfor Signed Numbers - Magnitude Comparator - Decoders / Encoders - <br> Multiplexers / Demultiplexers- Boolean Function Implementation using Multiplexers. | 9 |  |
| UNIT-III | 9 |  |  |
| Sequential Logic - Basic Latch - Flip-Flops (SR, D, JK, T and Master-Slave) - Triggering of Flip-Flops - Counters <br> - Design Procedure - Ripple Counters - BCD and Binary - Synchronous Counters. | 9 |  |  |
| UNIT-IV | 9 |  |  |
| Registers - Shift Registers - Registers with Parallel Load - Memory Unit - Examples of RAM, ROM, PROM, <br> EPROM - Reduction of State and Flow Tables - Race-Free State Assignment - Hazards. | 9 |  |  |

## Text Books

" Digital Design" By M. Morris Mano, , Prentice Hall of India

## Digital Systems

- It is a system that manipulates (processes) discrete elements of information.
- A digital system is a discrete information processing system.
- Examples of digital systems: - the calculator, - digital voltmeters, - and general-purpose computers.
- In such systems, all quantities are represented using two values.

Discrete elements of information are represented in a digital system by physical quantities called signals.

## Signal

Signal can be defined as a physical quantity, which contains some information. It is a function of one or more than one independent variables. Signals are of two types.

- Analog Signal
- Digital Signal


## Analog Signal

An analog signal is defined as the signal having continuous values. Analog signal can have infinite number of different values.
Examples of the analog signals:
Temperature, Pressure, Sound, Voltage, Current, Power

## Graphical representation of Analog Signal (Temperature)



Analog Signal
The circuits that process the analog signals are called as analog circuits or system. Examples of the analog system:

- Filter
- Amplifiers
- Television receiver
- Motor speed controller

Disadvantage of Analog Systems

- Less accuracy
- More noise effect
- More distortion
- More effect of weather


## Digital Signal

- A digital signal is defined as the signal which has only a finite number of distinct values.
- In the digital electronic calculator, the input is given with the help of switches. This input is converted into electrical signal which have two discrete values or levels.
- One of these may be called low level and another is called high level. The signal will always be one of the two levels. This type of signal is called digital signal.
Examples of the digital signal:
- Binary Signal
- Octal Signal
- Hexadecimal Signal


## Graphical representation of the Digital Signal (Binary)



The circuits that process the digital signals are called digital systems or digital circuits.
Examples of the digital systems:
Registers, Flip-flop, Counters, Microprocessors
Advantage of Digital Systems

- More accuracy
- Less distortion
- Easy communicate


## Comparison of Analog and Digital Signal

| S.No. | Analog Signal | Digital Signal |
| :--- | :--- | :--- |
| 1 | Analog signal has infinite <br> values. | Digital signal has a finite <br> number of values. |
| 2 | Analog signal has a <br> continuous nature. | Digital signal has a discrete <br> nature. |
| 3 | Analog signal is generated <br> by transducers and signal <br> generators. | Digital signal is generated <br> by A to D converter. |
| 4 | Example of analog signal - <br> sine wave, triangular waves. | Example of digital signal - <br> binary signal. |

## Merits/Advantages of Digital Systems

- Digital Systems are generally easier to Design.
- Digital Information storage is easy.
- Digital circuits are less affected by the noise.
- More Circuitry can be fabricated on integrated Circuits (ICs) chips.
- Display of data and other information is very convenient, accurate and elegant using digital techniques.


## Limitation of Digital Systems

- Only Analog Signal is available in the real world..
- An extra hardware is required to convert analog signal to digital signal.


## By Using

* Digital to Analog converter (DAC)
* Analog to digital converter (ADC)

A digital system can understand positional number system only where there are a few symbols called digits and these symbols represent different values depending on the position they occupy in the number.

## Number System

Number system is the language of digital system consisting a set of symbols called digits with rules defined for mathematical operations

* Non Positional Number System
* Positional Number System
- A number is formed by collection of digits.
- It has two components Integer and Fraction
- Both are separated by decimal point

Example:(467.34) ${ }_{10}$

## Non-positional Number Systems

Characteristics

- In non-positional number system , each symbol represents the same value regardless of its position in the number
- Use symbols such as I for 1 , II for 2, III for 3, IIII for 4, IIIII for 5, etc
- The symbols are simply added to find out the value of a particular number
Difficulty
- It is difficult to perform arithmetic with sucha number system


## Positional Number Systems

- In positional number system, each symbol represents different value depending on the position they occupy in a number.
- In positional number system, each system has a value that relates to the number directly next to it. The total value of a positional number is the total of the resultant value of all positions.

Example: 12 can be $1 \times 10+2 \times 1,10+2=12$.

## The value of each digit is determined by:

1. The digit itself
2. The position of the digit in the number
3. The base of the number system
(base $=$ total number of digits in the number system)
The maximum value of a single digit is always equal to one less than the valueof the base

## Decimal Number System

A positional number system

- Decimal number system has base 10 as it uses 10 symbols or digits ( $0,1,2,3,4,5,6,7,8,9$ ).
- The maximum value of a single digit is 9 (one less than the value of the base)
- Each position of a digit represents a specific power of the base (10)
- We use this number system in our day-to-day life

Example

$$
\begin{aligned}
2586_{10}= & \left(2 \times 10^{3}\right)+\left(5 \times 10^{2}\right)+\left(8 \times 10^{1}\right)+\left(6 \times 10^{0}\right) \\
& =2000+500+80+6
\end{aligned}
$$

## Decimal Number System

- Base (also called radix) $=10$
- 10 digits $\{0,1,2,3,4,5,6,7,8,9\}$
- Digit Position
- Integer \& fraction
- Digit Weight
- Weight $=(\text { Base })^{\text {Position }}$
- Magnitude

- Sum of "Digit x Weight"
- Formal Notation



## Octal Number System

A positional number system

- Octal number system has total 8 symbols or digits ( $0,1,2,3,4,5$, 6,7 ). Hence, its base $=8$
- The maximum value of a single digit is 7 (one less than the value of the base)
- Each position of a digit represents a specific power of the base (8)
- Since there are only 8 digits, 3 bits $\left(2^{3}=8\right)$ sufficient to represent any octal number in binary

Example

$$
\begin{aligned}
& 2057_{8}=\left(2 \times 8^{3}\right)+\left(0 \times 8^{2}\right)+(5 \times 81)+\left(7 \times 8^{0}\right) \\
& =1024+0+40+7 \\
& =1071_{10}
\end{aligned}
$$

## Octal Number System

- Base $=8$
- 8 digits $\{0,1,2,3,4,5,6,7\}$
- Weights
- Weight $=(\text { Base })^{\text {Position }}$
- Magnitude
- Sum of "Digit x Weight"
- Formal Notation

$$
\begin{array}{ccccc}
64 & 8 & 1 & 1 / 8 & 1 / 64 \\
\square & \square & \square & \square & \square \\
2 & 1 & 0 & -1 & -2 \\
5{ }^{*} 8^{2}+1 & { }^{*} 8^{1}+2{ }^{*} 8^{0}+7{ }^{*} 8^{-1}+4{ }^{*} 8^{-2} \\
= & (330.9375)_{10}
\end{array}
$$

## Binary Number System

A positional number system

- Binary number system has base 2 as it uses 2 digits from 0 to 1 .
- The maximum value of a single digit is 1 (one less than the value of the base)
- Each position of a digit represents a specific power of the base (2)
- This number system is used in computers


## Example

$$
\begin{aligned}
& 10101_{2}=\left(1 \times 2^{4}\right)+\left(0 \times 2^{3}\right)+\left(1 \times 2^{2}\right)+\left(0 \times 2^{1}\right) \times\left(1 \times 2^{0}\right) \\
& =16+0+4+0+1 \\
& =21_{10}
\end{aligned}
$$

## Binary Number System

Binary number system has base 2 as it uses 2 digits from 0 to 1 .

- Base $=2$
- 2 digits $\{0,1\}$, called binary dig $/$ ts or "bits"
- Weights
- Weight $=(\text { Base })^{\text {Position }}$
- Magnitude
- Sum of "Bit x Weight"
- Formal Notation
- Groups of bits 4 bits = Nibble

$$
8 \text { bits = Byte }
$$



11000101

## Hexadecimal Number System

A positional number system

- Hexadecimal number system has total 16 symbols or digits (0, 1, 2, 3, $4,5,6,7,8,9, A, B, C, D, E, F) . \quad$ Hence its base $=16$
- The symbols A, B, C, D, E and F represent the decimal values 10,11 , $12,13,14$ and 15 respectively
- The maximum value of a single digit is 15 (one less than the value of the base)
- Each position of a digit represents a specific power of the base (16)
- Since there are only 16 digits, 4 bits $\left(2^{4}=16\right)$ are sufficient to represent any hexadecimal number in binary

$$
\text { Ex. } \quad \begin{aligned}
\mathrm{AF}_{16} & =\left(1 \times 16^{2}\right)+(A \times 161)+(F \times 160) \\
& =1 \times 256+10 \times 16+15 \times 1 \\
& =256+160+15 \\
& =431_{10}
\end{aligned}
$$

## Hexadecimal Number System

- Uses 10 digits and 6 letters, 0,1,2,3,4,5,6,7,8,9,A,B,C,D,E,F.
- Base = 16
- 16 digits $\{0,1,2,3,4,5,6,7,8,9, A, B, C, D, E, F\}$
- Weights
- Weight $=(\text { Base })^{\text {Position }}$
- Magnitude
- Sum of "Digit x Weight"
- Formal Notation

$$
\begin{gathered}
2 \\
{ }^{*} 16^{2}+14 * 16^{1}+5^{*} 16^{0}+7 * 16^{-1}+10 * 16^{-2} \\
=(\mathbf{4 8 5 . 4 7 6 5 6 2 5})_{10}
\end{gathered}
$$

$(1 E 5.7 A)_{16}$

## Addition

- Decimal Addition



## Addition

$>$ Add and multiply the following numbers in the given base without converting to decimal.
(a) $(1230)_{4}$ and $(23)_{4}$

| 1 |
| ---: |
| 1230 |
| $+\quad 23$ |
| 1313 |


| 111 |
| :---: |
| 22 |
| 1230 |
| $\mathbf{x} \quad 23$ |
| 11010 |
| 3120 |
| 102210 |

(b) $(135.4)_{6}$ and (43.2) $)_{6}$
(c) $\mathbf{( 3 6 7 )}_{8}$ and $(715)_{8}$

## Binary Addition

- Column Addition



## Binary Subtraction

- Borrow a "Base" when needed



## Binary Multiplication

- Bit by bit



## Decimal (Integer) to Binary Conversion

- Divide the number by the 'Base' (=2)
- Take the remainder (either 0 or 1 ) as a coefficient
- Take the quotient and repeat the division

Example: (13) ${ }_{10}$ Quotient Remainder Coefficient

| $\mathbf{1 3 / 2}=$ | $\mathbf{6}$ | 1 | $\mathbf{a}_{\mathbf{0}}=1$ |
| ---: | :--- | :--- | :--- |
| $\mathbf{6} / \mathbf{2}=$ | $\mathbf{3}$ | 0 | $\mathbf{a}_{\mathbf{1}}=0$ |
| $\mathbf{3 / 2}=$ | $\mathbf{1}$ | 1 | $\mathbf{a}_{\mathbf{2}}=1$ |
| $\mathbf{1 / 2}=$ | $\mathbf{0}$ | 1 | $\mathbf{a}_{\mathbf{3}}=1$ |
| Answer: | $(13)_{10}=\left(\mathbf{a}_{3} a_{2} a_{1} a_{0}\right)_{2}=(1101)_{2}$ |  |  |

MSB
LSB

## Decimal (Fraction) to Binary Conversion

- Multiply the number by the 'Base’ (=2)
- Take the integer (either 0 or 1) as a coefficient
- Take the resultant fraction and repeat the division

Example: (0.625) ${ }_{10}$
Integer
Fraction
Coefficient

| 0.625 | * $\mathbf{2}=1$ | 25 | $\mathbf{a}_{-1}=1$ |
| :---: | :---: | :---: | :---: |
| 0.25 | * $\mathbf{2}=0$ | 5 | $\mathbf{a}_{-2}=0$ |
| 0.5 | * $\mathbf{2}=1$ | 0 | $\mathbf{a}_{-3}=1$ |
| Answer: | (0.625) | 0.a | $\left.\mathrm{a}_{-3}\right)_{2}=(0.10$ |

MSB LSB

## Decimal to Octal Conversion

Example: (175)

|  | Quotient | Remainder | Coefficient |
| :---: | :---: | :---: | :---: |
| $175 / 8=$ | 21 | 7 | $\mathbf{a}_{0}=7$ |
| $21 / 8=$ | 2 | 5 | $\mathbf{a}_{1}=5$ |
| $2 / 8=$ | 0 | 2 | $\mathbf{a}_{2}=2$ |
| Answer: $\quad(175)_{10}=\left(a_{2} a_{1} a_{0}\right)_{8}=(257)_{8}$ |  |  |  |

Example: (0.3125) ${ }_{10}$

|  | Integer | Fraction | Coefficient |
| :---: | :---: | :---: | :---: |
| 0.3125 | * $8=2$ | 5 | $\mathrm{a}_{-1}=$ |
| 0.5 | * 8 = 4 | 0 | $\mathbf{a}_{-2}=4$ |
| Answer: | $(0.3125)_{10}$ | .a-1 | $\left.\mathrm{a}_{-3}\right)_{8}=(0.24)_{8}$ |

## Binary - Octal Conversion

- $8=2^{3}$
- Each group of 3 bits represents an octal digit

Example:


| Octal | Binary |
| :---: | :---: |
| 0 | 000 |
| 1 | 001 |
| 2 | 010 |
| 3 | 011 |
| 4 | 100 |
| 5 | 110 |
| 7 | 111 |

Works both ways (Binary to Octal \& Octal to Binary)

## Binary - Hexadecimal Conversion

- $16=2^{4}$
- Each group of 4 bits represents a hexadecimal digit

Example:


| Hex | Binary |
| :---: | :---: |
| 0 | 0000 |
| 1 | 0001 |
| 2 | 0010 |
| 3 | 0011 |
| 4 | 0100 |
| 5 | 0101 |
| 6 | 0110 |
| 7 | 0111 |
| 8 | 1000 |
| 9 | 1001 |
| A | 1010 |
| B | 1011 |
| C | 1100 |
| D | 1101 |
| E | 1110 |
| $F$ | 1111 |

Works both ways (Binary to Hex \& Hex to Binary)

## Octal - Hexadecimal Conversion

- Convert to Binary as an intermediate step


## Example:



Works both ways (Octal to Hex \& Hex to Octal)

## Converting a Number of Some Base to a Number of Another Base

## Method

Step 1: Convert the original number to a decimal number (base 10)
Step 2: Convert the decimal number so obtained to the new base number

Example

$$
545_{6}=?_{4}
$$

Solution:
Step 1: Convert from base 6 to base 10

$$
\begin{aligned}
545_{6}=5 \times 6^{2} & +4 \times 6^{1}+5 \times 60 \\
& =5 \times 36+4 \times 6+5 \times 1 \\
& =180+24+5 \\
& =209_{10}
\end{aligned}
$$

Step 2: Convert 209 ${ }_{10}$ to base 4

4 | 209 | Remainders |
| :--- | ---: |
|  | $\mathbf{5 2}$ |
| 13 | $\mathbf{1}$ |
| 3 | $\mathbf{0}$ |
| 0 | $\mathbf{1}$ |
|  | $\mathbf{3}$ |

Hence, $\mathbf{2 0 9}_{10}=\mathbf{3 1 0 1}_{4}$
So, $545{ }_{6}=\mathbf{2 0 9}_{10}=\mathbf{3 1 0 1}_{4}$
Thus, $\mathbf{5 4 5}_{\mathbf{6}}=\mathbf{3 1 0 1}_{\mathbf{4}}$

Convert the following numbers with the indicated bases to decimal : $(4310)_{5}$, and (198) ${ }_{12}$

- $(4310)_{5}=0 \times 5^{0}+1 \times 5^{1}+3 \times 5^{2}+4 \times 5^{3}=0+5+75+500$
$(4310)_{5}=(580)_{10}$
- $(198))_{12}=8 \times 12^{0}+9 \times 12^{1}+1 \times 12^{2}=8+108+144$

$$
(198)_{12}=(260)_{10}
$$

## cimal, Binary, Octal and Hexadecimal

| Decimal | Binary | Octal | Hex |
| :---: | :---: | :---: | :---: |
| 00 | 0000 | 00 | 0 |
| 01 | 0001 | 01 | 1 |
| 02 | 0010 | 02 | 2 |
| 03 | 0011 | 03 | 3 |
| 04 | 0100 | 04 | 4 |
| 05 | 0101 | 05 | 5 |
| 06 | 0110 | 06 | 6 |
| 07 | 0111 | 07 | 7 |
| 08 | 1000 | 10 | 8 |
| 09 | 1001 | 11 | 9 |
| 10 | 1010 | 12 | A |
| 11 | 1011 | 13 | B |
| 12 | 1100 | 14 | C |
| 13 | 1101 | 15 | D |
| 14 | 1110 | 16 | E |
| 15 | 1111 | 17 | F |

### 1.5 Complements

- There are two types of complements for each base-r system: the radix complement and diminished radix complement.
- Diminished Radix Complement - (r-1)'s Complement
- Given a number $N$ in base $r$ having $n$ digits, the ( $r-1$ )'s complement of $N$ is defined as:

$$
\left(r^{n}-1\right)-N
$$

- Example for 6-digit decimal numbers:
- 9's complement is $\left(r^{n}-1\right)-N=\left(10^{6}-1\right)-N=999999-N$
- 9's complement of 546700 is 999999-546700 $=453299$
- Example for 7-digit binary numbers:
- 1 's complement is $\left(r^{n}-1\right)-N=\left(2^{7}-1\right)-N=1111111-N$
- 1 's complement of 1011000 is $1111111-1011000=0100111$
- Observation:
- Subtraction from $\left(r^{n}-1\right)$ will never require a borrow
- Diminished radix complement can be computed digit-by-digit
- For binary: $1-0=1$ and $1-1=0$


## Complements

- 1's Complement (Diminished Radix Complement)
- All '0’s become '1’s
- All '1's become '0’s

Example (10110000) ${ }_{2}$

$$
\Rightarrow(01001111)_{2}
$$

If you add a number and its 1 's complement ...

$$
\begin{array}{r}
10110000 \\
+\quad 01001111 \\
\hline \mathbf{1 1 1 1 1 1 1 1}
\end{array}
$$

## Complements

- Radix Complement

> The r's complement of an $n$-digit number $N$ in base $r$ is defined as $r^{n}-N$ for $N \neq 0$ and as 0 for $N=0$. Comparing with the $(r-1)$ 's complement, we note that the $r^{\prime}$ s complement is obtained by adding 1 to the $(r-1)$ 's complement, since $r^{n}-N=\left[\left(r^{n}-1\right)-N\right]+1$.

- Example: Base-10

The 10 's complement of 012398 is $\mathbf{9 8 7 6 0 2}$
The 10 's complement of 246700 is 753300

- Example: Base-2

The 2's complement of $\mathbf{1 1 0 1 1 0 0}$ is $\mathbf{0 0 1 0 1 0 0}$
The 2's complement of 0110111 is 1001001

## Complements

- 2's Complement (Radix Complement)
- Take 1's complement then add 1

OR - Toggle all bits to the left of the first ' 1 ' from the right

## Example:

Number:
10110000
10110000
1's Comp.:
01001111


01010000
01010000

## Complements

- Subtraction with Complements
- The subtraction of two $n$-digit unsigned numbers $M-N$ in base $r$ can be done as follows:

1. Add the minuend $M$ to the $r$ 's complement of the subtrahend $N$. Mathematically, $M$ $+\left(r^{n}-N\right)=M-N+r^{n}$.
2. If $M \geqq N$, the sum will produce and end carry $r^{n}$, which can be discarded; what is left is the result $M-N$.
3. If $M<N$, the sum does not produce an end carry and is equal to $r^{n}-(N-M)$, which is the $r$ 's complement of $(N-M)$. To obtain the answer in a familiar form, take the $r$ 's complement of the sum and place a negative sign in front.

## Complements

- Example 1.5
- Using 10's complement, subtract 72532-3250.

| $M$ | $=72532$ |
| ---: | :--- |
| 10's complement of $N$ | $=\frac{+96750}{169282}$ |
| Sum | $=$ |
| Discard end carry $10^{5}$ | $=\frac{-100000}{}$ |
| Answer | $=\quad 69282$ |

- Example 1.6
- Using 10's complement, subtract 3250-72532.

| M's complement of | $=$ | 03250 |
| ---: | :--- | ---: |
| N | $=$ | +27468 |
|  | $\mathrm{Sum}=$ | 30718 |

## There is no end carry.

## complements

- Example 1.7
- Given the two binary numbers $X=1010100$ and $Y=1000011$, perform the subtraction (a) $X-Y$; and (b) $Y-X$, by using 2 's complement.
(a) $\mathrm{X}=1010100$

2's complement of $Y=+0111101$

$$
\text { Sum }=10010001
$$

Discard end carry $2^{7}=-10000000$

$$
\text { Answer. } \mathrm{X}-\mathrm{Y}=0010001
$$

(b) \begin{tabular}{rrr}
Y \& $=1000011$ <br>

2's complement of X \& $=+$| 0101100 |
| :--- |
| Sum | \& 1101111 <br>

\&
\end{tabular}

There is no end carry. Therefore, the answer is Y -$X=-$ ( 2 's complement of $1101111)=-0010001$.

## omplements

- Subtraction of unsigned numbers can also be done by means of the $(r-1$ )'s complement. Remember that the $(r-1)$ 's complement is one less then the $r$ 's complement.
- Example 1.8
- Repeat Example 1.7, but this time using 1's complement.

| (a) $X-Y=1010100-1000011$ |  |  |
| :---: | :---: | :---: |
|  | $X=$ | 1010100 |
|  | 1's complement of $Y= \pm 0111100$ |  |
|  | Sum $=$ | 10010000 |
|  | End-around carry $=$ | 1 <br> $+\quad 1$ |
|  | Answer. $X-Y=$ | 0010001 |

$$
\text { (b) } \begin{array}{rlr|} 
& \\
Y-X=1000011-1010100 & \\
Y & =1000011 \\
\text { 1's complement of } X & =+0101011 \\
\text { Sum } & =1101110
\end{array}
$$

There is no end carry,
Therefore, the answer is $Y-X$
$=-(1$ 's complement of 1101110) $=-0010001$.

## Signed Binary Numbers

- To represent negative integers, we need a notation for negative values.
- It is customary to represent the sign with a bit placed in the leftmost position of the number since binary digits.
- The convention is to make the sign bit 0 for positive and 1 for negative.
- Example:

| Signed-magnitude representation: | 10001001 |
| :--- | :--- |
| Signed-1's-complement representation: | 11110110 |
| Signed-2's-complement representation: | 11110111 |

- Table 1.3 lists all possible four-bit signed binary numbers in the three representations.


## Signed Binary Numbers

## Table 1.3 <br> Signed Binary Numbers

| Decimal | Signed-2's <br> Complement | Signed-1's <br> Complement | Signed <br> Magnitude |
| :---: | :---: | :---: | :---: |
| +7 | 0111 | 0111 | 0111 |
| +6 | 0110 | 0110 | 0110 |
| +5 | 0101 | 0101 | 0101 |
| +4 | 0100 | 0100 | 0100 |
| +3 | 0011 | 0011 | 0011 |
| +2 | 0010 | 0010 | 0010 |
| +1 | 0001 | 0001 | 0001 |
| +0 | 0000 | 0000 | 0000 |
| -0 | - | 1111 | 1000 |
| -1 | 1111 | 1110 | 1001 |
| -2 | 1110 | 1101 | 1010 |
| -3 | 1101 | 1100 | 1011 |
| -4 | 1100 | 1011 | 1100 |
| -5 | 1011 | 1010 | 1101 |
| -7 | 1010 | 1001 | 1110 |
| -8 | 1001 | 1000 | 1111 |

## gned Binary Numbers

- Arithmetic addition
- The addition of two numbers in the signed-magnitude system follows the rules of ordinary arithmetic. If the signs are the same, we add the two magnitudes and give the sum the common sign. If the signs are different, we subtract the smaller magnitude from the larger and give the difference the sign if the larger magnitude.
- The addition of two signed binary numbers with negative numbers represented in signed-2's-complement form is obtained from the addition of the two numbers, including their sign bits.
- A carry out of the sign-bit position is discarded.
- Example:

| +66 | 00000110 | -6 | 11111010 |
| :--- | :--- | :---: | :--- |
| $\underline{+13}$ | $\underline{00001101}$ | $\underline{+13}$ | $\underline{00001101}$ |
| +19 | 00010011 | +7 | 00000111 |
| +6 | 00000110 | -6 | 11111010 |
| $\underline{-13}$ | $\underline{11110011}$ | $\underline{-13}$ | $\underline{11110011}$ |
| -7 | 11111001 | -19 | 11101101 |

## Signed Binary Numbers

- Arithmetic Subtraction
- In 2's-complement form:

1. Take the 2 's complement of the subtrahend (including the sign bit) and add it to the minuend (including sign bit).
2. A carry out of sign-bit position is discarded.

$$
\begin{aligned}
& ( \pm A)-(+B)=( \pm A)+(-B) \\
& ( \pm A)-(-B)=( \pm A)+(+B)
\end{aligned}
$$

- Example:

$$
(-6)-(-13)
$$

(11111010-11110011)
( $11111010+00001101$ )
00000111 (+7)

## Binary Codes

Digital data is represented, stored and transmitted as groups of binary digits also known as binary code.
Binary codes are codes which are represented in binary system with modification from the original ones.

$>$ Weighted codes: In weighted codes, each digit is assigned a specific weight according to its position. $>$ Non-weighted codes: Non weighted codes are codes that are not positionally weighted. That is, each position within the binary number is not assigned a fixed value..
$>$ Reflective codes: A code is reflective when the code is self complementing. In other words, when the code for $\mathbf{9}$ is the complement the code for $\mathbf{0 , 8} \mathbf{8}$ for $\mathbf{1 , 7}$ for $\mathbf{2 , 6}$ for $\mathbf{3}$ and $\mathbf{5}$ for $\mathbf{4}$.
$>$ Sequential codes: In sequential codes, each succeeding 'code is one binary number greater than its preceding code.
>Alphanumeric codes: Codes used to represent numbers, alphabetic characters, symbols
$>$ Error defecting and correcting codes: Codes which allow error defection and correction are called error detecting and' correcting codes.

Self-complementing codes 9's complement of a decimal number is obtained by interchanging all 1's and 0's by 0's and 1's respectively.

| Decimal | 2421 code | Excess code | 5211 code |
| :--- | :--- | :--- | :--- |
| 0 | 0000 | 0011 | 0000 |
| 1 | 0001 | 0100 | 0001 |
| 2 | 0010 | 0101 | 0011 |
| 3 | 0011 | 0110 | 0101 |
| 4 | 0100 | 0111 | 0111 |
| 5 | 1011 | 1000 | 1000 |
| 6 | 1100 | 1001 | 1010 |
| 7 | 1101 | 1010 | 1100 |
| 8 | 1110 | 1011 | 1110 |
| 9 | 1111 | 1100 | 1111 |

BCD Code

- A number with k decimal digits will require 4 k bits in BCD.
- Decimal 396 is represented in BCD with 12bits as 00111001 0110, with each group of 4 bits representing one decimal digit.
- A decimal number in BCD is the same as its equivalent binary number only when the number is between 0 and 9 .
- The binary combinations 1010 through 1111 are not used and have no meaning in BCD.

Table 1.4
Binary-Coded Decimal (BCD)

| Decimal <br> Symbol | BCD <br> Digit |
| :---: | :---: |
| 0 | 0000 |
| 1 | 0001 |
| 2 | 0010 |
| 3 | 0011 |
| 4 | 0100 |
| 5 | 0101 |
| 6 | 0110 |
| 7 | 0111 |
| 8 | 1000 |
| 9 | 1001 |

Example: Consider decimal 185 and its corresponding value in BCD and binary:
$(185)_{10}=(000110000101)_{\mathrm{BCD}}=(10111001)_{2}$

| 4 | 0100 | 4 | 0100 | 8 | 1000 |
| ---: | ---: | ---: | ---: | ---: | ---: |
| +5 | +0101 | $\frac{+8}{+1000}$ | $\underline{+9}$ | $\frac{+1001}{1100}$ | 17 |
| 9 | 1001 | 12 | 1001 |  |  |
|  |  | +0110 |  | +0110 |  |
|  |  | 10010 |  | 10111 |  |

## inary Codes

- Other Decimal Codes

Table 1.5
Four Different Binary Codes for the Decimal Digits

| Decimal <br> Digit | $\mathbf{B C D}$ <br> $\mathbf{8 4 2 1}$ | $\mathbf{2 4 2 1}$ | Excess-3 | $\mathbf{8 , 4 ,} \mathbf{- \mathbf { 2 } , \mathbf { - 1 }}$ |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0000 | 0000 | 0011 | 0000 |
| 1 | 0001 | 0001 | 0100 | 0111 |
| 2 | 0010 | 0010 | 0101 | 0110 |
| 3 | 0011 | 0011 | 0110 | 0101 |
| 4 | 0100 | 0100 | 0111 | 0100 |
| 5 | 0101 | 1011 | 1000 | 1011 |
| 6 | 0110 | 1100 | 1001 | 1010 |
| 7 | 0111 | 1101 | 1010 | 1001 |
| 8 | 1000 | 1110 | 1011 | 1000 |
| 9 | 1001 | 1111 | 1100 | 1111 |
|  | 1010 | 0101 | 0000 | 0001 |
| Unused | 1011 | 0110 | 0001 | 0010 |
| bit | 1100 | 0111 | 0010 | 0011 |
| combi- | 1101 | 1000 | 1101 | 1100 |
| nations | 1110 | 1001 | 1110 | 1101 |
|  | 1111 | 1010 | 1111 | 1110 |

## Gray Code:

The gray code belongs to a class of codes called minimum change codes, in which only one bit in the code changes when moving from one code to the next.

The Gray code is non-weighted code, as the position of bit does not contain any weight.

The gray code is a reflective digital code which has the special property that any two subsequent numbers codes differ by only one bit. This is also called a unitdistance code. In digital Gray code has got a special place.

- Gray Code
- The advantage is that only bit in the code group changes in going from one number to the next.
- Gray code is also known as reflected binary code.
- We call it gray code after frank gray.
- Unweighted code.
- Two successive values differ in only 1 bit.
- Binary number is converted to gray code to reduce switching operation.
- Unit distance code \& min. error code.
- Cyclic code
- Low power design.

Table 1.6
Gray Code

| Gray <br> Code | Decimal <br> Equivalent |
| :---: | :---: |
| 0000 | 0 |
| 0001 | 1 |
| 0011 | 2 |
| 0010 | 3 |
| 0110 | 4 |
| 0111 | 5 |
| 0101 | 6 |
| 0100 | 7 |
| 1100 | 8 |
| 1101 | 9 |
| 1111 | 10 |
| 110 | 11 |
| 1010 | 12 |
| 1011 | 13 |
| 1001 | 14 |
| 1000 | 15 |

## Binary Codes

## Binary to Gray code Conversion

Step 1: Record the MSB as it is.
Step 2: Add the MSB to the next bit, record the sum and neglect the carry. (X-OR)
Step 3: Repeat the process.
Exp. Convert 1011 to Gray code.
MSB LSB
1011 (Binary)
1110 (Gray)

## Binary Codes

## Gray code to Binary Conversion

Step 1: Record the MSB as it is.
Step 2: Add the MSB to the next bit of Gray code, record the sum and neglect the carry. (X-OR)
Step 3: Repeat the process.
Exp. Convert 1011 to Gray code.
MSB LSB
$\begin{array}{ll}1.201 \\ 1101 & \text { (Gray) } \\ 1101 & \text { (Binary) }\end{array}$

American Standard Code for Information Interchange (ASCII) Character Code

Table 1.7
American Standard Code for Information Interchange (ASCII)

| $b_{4} b_{3} b_{2} b_{1}$ | $b_{7} b_{6} b_{5}$ |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 000 | 001 | 010 | 011 | 100 | 101 | 110 | 111 |
| 0000 | NUL | DLE | SP | 0 | @ | P | - | p |
| 0001 | SOH | DC1 | ! | , | A | Q | a | q |
| 0010 | STX | DC2 | * | 2 | B | R | b | r |
| 0011 | ETX | DC3 | \# | 3 | C | S | c | s |
| 0100 | EOT | DC4 | \$ | 4 | D | T | d | $t$ |
| 0101 | ENQ | NAK | \% | 5 | E | U | e | u |
| 0110 | ACK | SYN | \& | 6 | F | V | f | $v$ |
| 0111 | BEL | ETB | . | 7 | G | W | g | w |
| 1000 | BS | CAN | ( | 8 | H | X | h | x |
| 1001 | HT | EM | ) | 9 | I | Y | i | y |
| 1010 | LF | SUB | * | : | J | Z | j | z |
| 1011 | VT | ESC | + | ; | K | [ | k | \{ |
| 1100 | FF | FS | . | $<$ | L | 1 | 1 | 1 |
| 1101 | CR | GS | - | $=$ | M | 1 | m | \} |
| 1110 | SO | RS | . | > | N | $\wedge$ | n | $\sim$ |
| 1111 | SI | US | 1 | ? | O | - | o | DEL |

ASCII Character Code

## Control characters

| NUL | Null | DLE | Data-link escape |
| :--- | :--- | :--- | :--- |
| SOH | Start of heading | DC1 | Device control 1 |
| STX | Start of text | DC2 | Device control 2 |
| ETX | End of text | DC3 | Device control 3 |
| EOT | End of transmission | DC4 | Device control 4 |
| ENQ | Enquiry | NAK | Negative acknowledge |
| ACK | Acknowledge | SYN | Synchronous idle |
| BEL | Bell | ETB | End-of-transmission block |
| BS | Backspace | CAN | Cancel |
| HT | Horizontal tab | EM | End of medium |
| LF | Line feed | SUB | Substitute |
| VT | Vertical tab | ESC | Escape |
| FF | Form feed | FS | File separator |
| CR | Carriage return | GS | Group separator |
| SO | Shift out | RS | Record separator |
| SI | Shift in | US | Unit separator |
| SP | Space | DEL | Delete |

## SCII Character Codes and Properties

- American Standard Code for Information Interchange (Refer to Table 1.7)
- A popular code used to represent information sent as character-based data.
- It uses 7-bits to represent:
- 94 Graphic printing characters.
- 34 Non-printing characters.
- Some non-printing characters are used for text format (e.g. BS = Backspace, CR = carriage return).
- Other non-printing characters are used for record marking and flow control (e.g. STX and ETX start and end text areas).
- ASCII has some interesting properties:
- Digits 0 to 9 span Hexadecimal values $30_{16}$ to $39_{16}$
- Upper case A-Z span $41_{16}$ to $5 \mathrm{~A}_{16}$
- Lower case a-z span $61_{16}$ to $7 \mathrm{~A}_{16}$
- Lower to upper case translation (and vice versa) occurs by flipping bit 6.


## Error-Detecting Code

- Data can be corrupted during transmission. For reliable communication, errors must be detected and corrected.
- To detect errors in data communication and processing, an eighth bit is sometimes added to the ASCII character to indicate its parity.
- A parity bit is an extra bit included with a message to make the total number of 1 's either even or odd.
- Example:
- Consider the following two characters and their even and odd parity:

|  | With even parity | With odd parity |
| :--- | :---: | :---: |
| ASCII A $=1000001$ | 01000001 | 11000001 |
| ASCII T $=1010100$ | 11010100 | 01010100 |

## Error-Detecting Code

- Redundancy (e.g. extra information), in the form of extra bits, can be incorporated into binary code words to detect and correct errors.
- A simple form of redundancy is parity, an extra bit appended onto the code word to make the number of 1's odd or even. Parity can detect all single-bit errors and some multiple-bit errors.
- A code word has even parity if the number of 1's in the code word is even.
- A code word has odd parity if the number of 1 's in the code word is odd.
- Example: Message A: 100010011 (even parity)

Message B: 100010010 (odd parity)

## Hamming Codes

- It is not only provide to detect an error but also find out where the error occurred and correct it.
- In this code the receiver can detect whether there is any error in the received signal or not and also the bit (position) at which error happened.
- The formula to calculate the number of bits of hamming code is given as (number of parity bits depend on the message)

$$
m+k+1 \leq 2^{k}
$$

Where $m$ number of message bits
$k$ number of parity bits required
Thus k gives the number of bits that are to transmitted along with message.

Hence we have to make the hamming code at the transmitted end.
Let us assume as a seven bit message: 1010110
hence $m=7$
So, putting $\mathrm{m}=7$ in the formula, the k is satisfied by

$$
\begin{aligned}
& \mathrm{k}=4 \\
& \quad 7+\mathrm{k}+1 \leq 2^{\mathrm{k}} \\
& \quad 7+4+1 \leq 2^{4} \\
& \\
& 11+1 \leq 16
\end{aligned}
$$

So, the number of hamming code bits required to be transmitted along with the message are 4 .

Hence, the positions of parity bits are given as:

Positions (parity bits) Checks (bits to be checked)
$\mathrm{p}_{8} \mathrm{p}_{4} \mathrm{p}_{2} \mathrm{p}_{1}$
00000
$\mathrm{P}_{1}=2^{0}$ ( ${ }^{\text {st }}$ Column)
(1,3,5,7,9,11) bits
$\mathrm{P}_{2}=2^{1}\left(2^{\text {nd }}\right.$ Column)
$\mathrm{P}_{4}=2^{2}$ ( 3rd $^{\text {rd }}$ Column)
$\mathrm{P}_{8}=2^{3}$ (44 Column)
(2,3,6,7,10,11) bits
$(4,5,6,7,9)$ bits
$(8,9,10,11)$ bits
Thus the message will now be 11 bits long $\begin{array}{lllllllllll}11 & 10 & 9 & 8 & 7 & 6 & 5 & 4 & 3 & 2 & 1\end{array}$ $\mathrm{m}_{11} \mathrm{~m}_{10} \mathrm{~m}_{9} \mathrm{P}_{8} \mathrm{~m}_{7} \mathrm{~m}_{6} \mathrm{~m}_{5} \mathrm{P}_{4} \mathrm{~m}_{3} \mathrm{P}_{2} \mathrm{P}_{1}$ $1 \begin{array}{llllllllll}1 & 0 & 1 & \mathrm{P}_{8} & 0 & 1 & 1 & \mathrm{P}_{4} & 0 & \mathrm{P}_{2} \mathrm{P}_{1}\end{array}$

The value of parity bits can be calculated as: Checks

Even Parity

$$
\begin{array}{r}
\mathrm{P}_{1} \rightarrow(1,3,5,7,9,11) \\
\mathrm{P}_{1} 01011
\end{array}
$$

$$
\begin{aligned}
& m_{11} m_{10} m_{9} P_{8} m_{7} m_{6} m_{5} P_{4} m_{3} P_{2} P_{1} \\
& 1 \begin{array}{llllllllll}
1 & 0 & 1 & \mathrm{P}_{8} & 0 & 1 & 1 & \mathrm{P}_{4} & 0 & \mathrm{P}_{2} \\
\mathrm{P}_{1}
\end{array} \\
& \text { Checks } \\
& \mathrm{P}_{2} \rightarrow(2,3,6,7,10,11) \\
& \mathrm{P}_{2} 011001 \\
& \mathrm{P}_{4} \rightarrow(4,5,6,7,9) \\
& \mathrm{P}_{4} 1101 \\
& \mathrm{P}_{8} \rightarrow(8,9,10,11) \\
& \mathrm{P}_{4} 101 \\
& \text { Even Parity } \\
& P_{2}=0 \\
& \mathrm{P}_{4}=1 \\
& \mathrm{P}_{8}=0 \\
& \text { The hamming's code is as } \\
& 10100111001 \\
& \mathrm{p}_{8} \mathrm{p}_{4} \mathrm{p}_{2} \mathrm{p}_{1} \\
& 00000 \\
& 10001 \\
& 20010 \\
& 30011 \\
& 40100 \\
& 50101 \\
& 60110 \\
& 70111 \\
& 81000 \\
& 9 \quad 1001 \\
& 101010 \\
& 111011 \\
& \text { Reception of hamming's code and error detection } \\
& \text { Let the received code for same message is as }
\end{aligned}
$$

```
1
m11 m}\mp@subsup{m}{10}{}\mp@subsup{m}{9}{}\mp@subsup{P}{8}{}\mp@subsup{m}{7}{}\mp@subsup{m}{6}{}\mp@subsup{m}{5}{\prime}\mp@subsup{P}{4}{}\mp@subsup{m}{3}{}\mp@subsup{P}{2}{}\mp@subsup{P}{1}{
```

here checks

Read from bottom to top side
$\mathrm{P}_{8} \mathrm{P}_{4} \mathrm{P}_{2} \mathrm{P}_{1} \Longrightarrow 0111=7^{\text {th }}$ bit
Hence the error is at $7^{\text {th }}$ bit
To eliminate the error always complement the error bit.

$$
10100111001
$$

If there is no error then $\mathrm{P}_{8} \mathrm{P}_{4} \mathrm{P}_{2} \mathrm{P}_{1} \rightarrow 0000$. which means there is no error.

## Binary Storage and Registers

- Registers
- A binary cell is a device that possesses two stable states and is capable of storing one of the two states.
- A register is a group of binary cells. A register with $n$ cells can store any discrete quantity of information that contains $n$ bits.


## n cells $\quad \rightarrow \quad \mathbf{2}^{\mathrm{n}}$ possible states

- A binary cell
- Two stable state
- Store one bit of information
- Examples: flip-flop circuits, capacitor
- A register
- A group of binary cells
- AX in x86 CPU
- Register Transfer
- A transfer of the information stored in one register to another.
- One of the major operations in digital system.
- An example in next slides.


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 General-Purpose Computers are an example of binary systems. A Digital Computer Example

## The Digital Computer

- The memory stores programs as well as input, output, ....
- The datapath (ALU) performs arithmetic and other data processing operations as specified by the program.
- The control unit supervises the flow of information between the various units.
- A datapath, when combined with the control unit, forms a component referred to as a central processing unit, or CPU.
- Input device: keyboard.
- Output device: LCD (liquid crystal display).
-These devices use digital logic circuits


## Transfer of information



## Transfer of information



- The other major component of a digital system
- Circuit elements to manipulate individual bits of information
- Load-store machine

| LD | R1; |
| :--- | :--- |
| LD | R2; |
| ADD | R3, R2, R1; |
| SD | R3; |

Figure 1.2 Example of binary information processing

## Binary Logic

## - Definition of Binary Logic

- Binary logic consists of binary variables and a set of logical operations.
- The variables are designated by letters of the alphabet, such as $A, B, C, x, y, z$, etc, with each variable having two and only two distinct possible values: 1 and 0 ,
- Three basic logical operations: AND, OR, and NOT.

1. AND: This operation is represented by a dot or by the absence of an operator. For example, $x \cdot y=z$ or $x y=z$ is read " $x$ AND $y$ is equal to $z$," The logical operation AND is interpreted to mean that $z=1$ if only $x=1$ and $y=1$; otherwise $z=0$. (Remember that $x, y$, and $z$ are binary variables and can be equal either to 1 or 0 , and nothing else.)
2. OR: This operation is represented by a plus sign. For example, $x+y=z$ is read " $x$ OR $y$ is equal to $z$," meaning that $z=1$ if $x=1$ or $y=1$ or if both $x=1$ and $y=1$. If both $x=0$ and $y=0$, then $z=0$.
3. NOT: This operation is represented by a prime (sometimes by an overbar). For example, $\mathrm{x}^{\prime}=\mathrm{z}$ ( or $\bar{x}=z$ ) is read "not $x$ is equal to $z$," meaning that $z$ is what $z$ is not. In other words, if $x=1$, then $z=0$, but if $x=0$, then $z=1$, The NOT operation is also referred to as the complement operation, since it changes a 1 to 0 and a 0 to 1.

Binary Logic Function

F(vars) = expression

## variables

-Operators ( $+,^{\bullet},{ }^{\bullet}$ ) -Variables
-Constants ( 0, 1 )
$\square$ Groupings (parenthesis)
Example: $F(a, b)=a^{\prime} \cdot b+b^{\prime}$

$$
G(x, y, z)=x \cdot\left(y+z^{\prime}\right)
$$

## Binary Logic Function

- AND
- OR
- NOT

- $F(a, b)=a \cdot b, \quad F$ is 1 if and only if $a=b=1$
- $G(a, b)=a+b, G$ is 1 if either $a=1$ or $b=1$
- $H(a)=a \prime, H$ is 1 if $a=0$


## Binary Logic gates

- Truth Tables, Boolean Expressions, and Logic Gates
- Truth table: tabular form that uniquely represents the relationship between the input variables of a function and its output


| $x$ | $y$ | $z$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 0 |
| 1 | 0 | 0 |
| 1 | 1 | 1 |

$$
\mathbf{z}=\mathbf{x} \cdot \mathbf{y}=\mathbf{x} \mathbf{y}
$$



| $x$ | $y$ | $z$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 1 |

$\mathbf{z}=\mathbf{x}+\mathbf{y}$
$\mathbf{z}=\overline{\mathbf{x}}=\mathbf{x}$,



Logic gates are abstractions of electronic circuit components that operate on one or more input signals to produce an output signal.


| $A$ | B | X |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 1 |



Inverter (NOT)


| Logic Function | Boolean Notation |
| :---: | :---: |
| AND | $A . B$ |
| $O R$ | $A+B$ |
| NOT | $\bar{A}$ |
| NAND | $\overline{A . B}$ |
| NOR | $\overline{A+B}$ |
| $E X-O R$ | $(A \bar{B} \bar{B})+(\bar{A} \cdot \bar{B})$ or $A \oplus B$ |
| EX-NOR | $(\bar{A} \bar{B} \bar{B})+\operatorname{Or} \bar{A} \bar{A} \bar{B}$ |

## Universal Gate

- NAND and NOR Gates are called Universal Gates because AND, OR and NOT gates can be implemented \&created by using these gates.

NAND Gate Implementations


NOR Gate Implementations

## inary Logic

- Logic gates
- Graphic Symbols and Input-Output Signals for Logic gates:

(a) Two-input AND gate
(b) Two-input OR gate
(c) NOT gate or inverter

Fig. 1.4 Symbols for digital logic circuits


Fig. 1.5 Input-Output signals for gates

## Binary Logic

- Logic gates
- Graphic Symbols and Input-Output Signals for Logic gates:

(a) Three-input AND gate
(b) Four-input OR gate

Fig. 1.6 Gates with multiple inputs

## Combinational Logic Circuit from Logic Function

- Consider function $F=A^{\prime}+B \cdot C^{\prime}+A^{\prime} \bullet B^{\prime}$
- A combinational logic circuit can be constructed to implement F, by appropriately connecting input signals and logic gates:
- Circuit input signals $\rightarrow$ from function variables (A, B, C)
- Circuit output signal $\rightarrow$ function output (F)
- Logic gates $\rightarrow$ from logic operations



## Combinational Logic Circuit from Logic Function

- In order to design a cost-effective and efficient circuit, we must minimize the circuit's size (area) and propagation delay (time required for an input signal change to be observed at the output line)
- Observe the truth table of $F=A^{\prime}+B \cdot C^{\prime}$ $+A^{\prime} \cdot B^{\prime}$ and $G=A^{\prime}+B \cdot C^{\prime}$
- Truth tables for F and G are identical $\rightarrow$ same function
- Use G to implement the logic circuit (less components)


## Combinational Logic Circuit from Logic Function



## Boolean Algebra

## Boolean Algebra : George Boole(English mathematician), 1854

- It was Invented by George Boole in year 1854
- An algebraic structure defined by a set $B=\{0,1\}$, together with two binary operators (+ and $\cdot$ ) and a unary operator ( )
"An Investigation of the Laws of Thought, on Which Are Founded the Mathematical Theories of Logic and Probabilities"


## Boolean Algebra

 $\{(1,0), \operatorname{Var},(N O T, ~ A N D, ~ O R)\}$> Boolean algebra is a useful tool for simplifying digital circuits.
$>$ Terminology:
$\checkmark$ Literal: A variable or its complement
$\checkmark$ Product term: literals connected by •
$\checkmark$ Sum term: literals connected by +

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## Basic Functions

Boolean functions: NOT, AND, OR, exclusive OR(XOR) : odd function exclusive NOR(XNOR) : even function(equivalence) Boolean functions for (a) AND, (b) OR, (c) XOR, and (d) NOT

| $x$ | $y$ | $x \wedge y$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 0 |
| 1 | 0 | 0 |
| 1 | 1 | 1 |

(a)

| $x$ | $y$ | $x \vee y$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 1 |

(b)

| $x$ | $y$ | $x \oplus y$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 0 |

(c)

(d)

## Basic functions

- AND $\mathbf{Z}=\mathbf{X} \cdot \mathbf{Y}$ or $\mathbf{Z}=\mathbf{X Y}$
$Z=1$ if and only if $X=1$ and $Y=1$, otherwise $Z=0$
- OR $Z=X+Y$
$Z=1$ if $X=1$ or if $Y=1$, or both $X=1$ and $Y=1 . Z=0$ if and only if $X=0$ and $Y=0$
- NOT $Z=X^{\prime}$ or
$\mathrm{Z}=1$ if $\mathrm{X}=0, \mathrm{Z}=0$ if $\mathrm{X}=1$

Boolean functions for (a) NAND, (b) NOR, and (c) XNOR

| $x$ | $y$ | NAND |
| :---: | :---: | :---: |
| 0 | 0 | 1 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 0 |

(a)

(b)

(c)

All possible binary boolean functions

| $x$ | $y$ | 0 | $\wedge$ | $x y^{\prime}$ | $x$ | $x^{\prime} y$ | $y$ | $\oplus$ | $\vee$ | NOR | XNOR | $y^{\prime}$ | $x+y^{\prime}$ | $x^{\prime}$ | $x^{\prime}+y$ | NAND | 1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| 0 | 1 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |
| 1 | 0 | 0 | 0 | 1 | 1 | 0 | 0 | 1 | 1 | 0 | 0 | 1 | 1 | 0 | 0 | 1 | 1 |
| 1 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 |

## Basic Identities of Boolean Algebra

| 1. | $X+0=X$ |  | $X \cdot 1=X$ | The relationship between a single variable $\mathbf{X}$, its complement $\mathrm{X}^{\prime}$, and the binary constants 0 and 1 |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 3. | $X+1=1$ | 4. | $X \cdot 0=0$ |  |  |
| 5. | $X+X=X$ | 6. | $X \cdot X=X$ |  |  |
| 7. | $X+\bar{X}=1$ | 8. | $X \cdot \bar{X}=0$ |  |  |
| 9. | $\overline{\bar{X}}=X$ |  |  |  |  |
| 10. | $X+Y=Y+X$ | 11. | $X Y=Y X$ |  | Commutative |
| 12. | $X+(Y+Z)=(X+Y)+Z$ | 13. | $X(Y Z)=(X Y)$ |  | Associative |
| 14. | $X(Y+Z)=X Y+X Z$ | 15. | $X+Y Z=(X+$ | $Y)(X+Z)$ | Distributive |
| 16. | $\overline{X+Y}=\bar{X} \cdot \bar{Y}$ | 17. | $\overline{X \cdot Y}=\bar{X}+\bar{Y}$ |  | DeMorgan's |

Duality: The dual of an expression is obtained by exchanging ( $\bullet$ and + ), and ( 1 and 0 ) in it, provided that the precedence of operations is not changed.

## Laws of Boolean Algebra

- Commutative Law: the order of literals does not matter


$$
A B=B A
$$



- Associative Law: the grouping of literals does not matter

$$
A+(B+C)=(A+B)+C(=A+B+C)
$$

$$
A(B C)=(A B) C(=A B C)
$$





## Rules of Boolean Algebra

$A+0=A \quad$ In math if you add 0 you have changed nothing in Boolean Algebra ORing with 0 changes nothing
$A \cdot 0=0 \quad$ In math if 0 is multiplied with anything you get 0 . If you AND anything with 0 you get 0
$\checkmark A \cdot 1=A \quad$ ANDing anything with 1 will yield the anything
$\checkmark \quad A+A=A \quad$ ORing with itself will give the same
$\checkmark A+A^{\prime}=1 \quad$ Either $A$ or $A^{\prime}$ must be 1 so $A+A^{\prime}=1$
$\checkmark A \cdot A=A$ ANDing with itself will give the same result
$\checkmark A \cdot A^{\prime}=0 \ln$ digital Logic $1^{\prime}=0$ and $0^{\prime}=1$, so $A A^{\prime}=0$ since one of the inputs must be 0 .
$\checkmark A=\left(A^{\prime}\right)^{\prime}$ If you not something twice you are back to the beginning

## $\checkmark \quad A+A^{\prime} B=A+B$

If $A$ is 1 the output is 1 If $A$ is 0 the output is $B$
$\checkmark A+A B=A$
$\checkmark(A+B)(A+C)=A+B C$

## DeMorgan's Theorem

$-(A \bullet B)^{\prime}=A^{\prime}+B^{\prime}$ and $(A+B)^{\prime}=A^{\prime} \bullet B^{\prime}$

- DeMorgan's theorem will help to simplify digital circuits using NORs and NANDs his theorem states



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## Boolean Analysis of Logic Circuits

- Constructing a Truth Table for a Logic Circuit

- Convert the expression into the min-terms containing all the input literals
- Get the numbers from the min-terms
- Putting ' 1 's in the rows corresponding to the min-terms and ' 0 's in the remains

$$
\text { Ex) } \begin{aligned}
A(B+C D)=A B\left(C+C^{\prime}\right)\left(D+D^{\prime}\right)+A\left(B+B^{\prime}\right) C D=A B C\left(D+D^{\prime}\right)+A B C^{\prime}\left(D+D^{\prime}\right)+A B C D+A B^{\prime} C D \\
=A B C D+A B C D^{\prime}+A B C^{\prime} D+A B C^{\prime} D^{\prime}+A B C D+A B^{\prime} C D=A B C D+A B C D^{\prime}+A B C^{\prime} D+A B C^{\prime} D^{\prime} \\
+A B^{\prime} C D=m 12+m 13+m 14+m 15=\Sigma(11,12,13,14,15)
\end{aligned}
$$

$$
A(B+C D)=m 11+m 12+m 13+m 14+m 15=\Sigma(11,12,13,14,15)
$$

| Input |  |  |  | Output |
| :---: | :---: | :---: | :---: | :---: |
| A | B | C | D | $A(B+C D)$ |
| O | O | O | O | 0 |
| 0 | 0 | 0 | 1 | 0 |
| O | O | 1 | 0 | 0 |
| 0 | 0 | 1 | 1 | 0 |
| 0 | 1 | 0 | 0 | 0 |
| O | 1 | O | 1 | 0 |
| 0 | 1 | 1 | 0 | 0 |
| 0 | 1 | 1 | 1 | 0 |
| 1 | 0 | 0 | 0 | 0 |
| 1 | 0 | 0 | 1 | 0 |
| 1 | 0 | 1 | 0 | 0 |
| 1 | O | 1 | 1 | 1 |
| 1 | 1 | 0 | 0 | 1 |
| 1 | 1 | 0 | 1 | 1 |
| 1 | 1 | 1 | O | 1 |
| 1 | 1 | 1 | 1 | 1 |

Any logic expression may be expressed in two standard forms:

- The Sum-of-Products(SOP) Form Ex) $A B+A B C, A B C+C D E+B^{\prime} C D^{\prime}$
- The Product-of-Sums(POS) Form Ex) $(A+B)(A+B+C),(A+B+C)(C+D+E)\left(B^{\prime}+C+D^{\prime}\right)$

DPrinciple of Duality : SOP $\Leftrightarrow$ POS ( change AND $\Leftrightarrow O R, 0 \Leftrightarrow 1$ )

- Domain of a Boolean Expression : The set of variables contained in the expression Ex) $A^{\prime} B+A B^{\prime} C$ : the domain is $\{A, B, C\}$


## $\checkmark$ Standard SOP Form (Canonical SOP Form)

- For all the missing variables, apply $\left(x+x^{\prime}\right)=1$ to the AND terms of the expression
- List all the min-terms in forms of the complete set of variables in ascending order

Ex : Convert the following expression into standard SOP form: $A B^{\prime} C+A^{\prime} B^{\prime}+A B C^{\prime} D$
Sol) domain $=\{A, B, C, D\}, A B^{\prime} C\left(D^{\prime}+D\right)+A^{\prime} B^{\prime}\left(C^{\prime}+C\right)\left(D^{\prime}+D\right)+A B C^{\prime} D$
$=A B^{\prime} C D^{\prime}+A B^{\prime} C D+A^{\prime} B^{\prime} C^{\prime} D^{\prime}+A^{\prime} B^{\prime} C^{\prime} D+A^{\prime} B^{\prime} C D^{\prime}+A^{\prime} B^{\prime} C D+A B C^{\prime} D$
$=1010+1011+0000+0001+0010+0011+1101=0+1+2+3+10+11+13=$ $\Sigma(0,1,2,3,10,11,13)$

## Standard POS Form (Canonical POS Form)

- For all the missing variables, apply $\left(x^{\prime} x\right)=0$ to the OR terms of the expression
- List all the max-terms in forms of the complete set of variables in ascending order

Ex : Convert the following expression into standard POS form:
$\left(A^{\prime}+B^{\prime}+C\right)\left(B^{\prime}+C+D^{\prime}\right)\left(A+B^{\prime}+C^{\prime}+D\right)$
Sol domain=\{A,B,C,D\}, (A+B'+C)(B'+C+D')(A+B'+C'+D) $=\left(\mathbf{A}+\mathrm{B}^{\prime}+\mathrm{C}^{\prime}+\mathrm{D}^{\prime} \mathrm{D}\right)\left(\mathrm{A}^{\prime} \mathbf{A}+\mathrm{B}^{\prime}+\mathrm{C}^{\prime}+\mathrm{D}^{\prime}\right)\left(\mathbf{A}+\mathrm{B}^{\prime}+\mathrm{C}^{\prime}+\mathrm{D}\right)$
$=\left(A+B^{\prime}+C+D^{\prime}\right)\left(A+B^{\prime}+C+D\right)\left(A^{\prime}+B^{\prime}+C^{\prime}+D^{\prime}\right)\left(A+B^{\prime}+C+D^{\prime}\right)\left(A+B^{\prime}+C\right.$
$\left.\left.{ }^{\prime}+D\right)=(0100)\right)(0101)(0110)(1101)=\Pi(4,5,6,13)$
$A+B C=(A+B)(A+C)$
Distributed Law

## Converting Standard SOP to Standard POS

Step 1. Evaluate each product term in the SOP expression. Determine the binary numbers that represent the product terms
Step 2. Determine all of the binary numbers not included in the evaluation in Step 1
Step 3. Write in equivalent sum term for each binary number Step 2 and expression in POS form

## Ex : Convert the following SOP to POS

Sol) SOP $=A^{\prime} B^{\prime} C^{\prime}+A^{\prime} B C^{\prime}+A^{\prime} B C+A B^{\prime} C+A B C=0+2+3+5+7=\Sigma(0,2,3,5,7)$
POS=(1)(4)(6) $=\Pi(1,4,6)\left(=\left(A+B+C^{\prime}\right)\left(A^{\prime}+B+C\right)\left(A^{\prime}+B^{\prime}+C\right)\right)$

- SOP and POS Observations
- Canonical Forms (Sum-of-minterms, Product-of-Maxterms), or other standard forms (SOP, POS) differ in complexity
- Boolean algebra can be used to manipulate equations into simpler forms
- Simpler equations lead to simpler implementations


## Definitions

- Literal: A variable or its complement
- Product term: literals connected by •
- Sum term: literals connected by +
- Minterm: a product term in which all the variables appear exactly once, either complemented or uncomplemented
- Maxterm: a sum term in which all the variables appear exactly once, either complemented or uncomplemented


## Minterms

- Represents exactly one combination in the truth table.
- Denoted by $m_{j}$, where $j$ is the decimal equivalent of the minterm's corresponding binary combination ( $b_{j}$ ).
- A variable in $m_{j}$ is complemented if its value in $b_{j}$ is 0 , otherwise is uncomplemented.
- Example: Assume 3 variables ( $\mathrm{A}, \mathrm{B}, \mathrm{C}$ ), and $j=3$. Then, $\mathrm{b}_{j}=011$ and its corresponding minterm is denoted by $m_{j}=A^{\prime} B C$


## Maxterm

- Represents exactly one combination in the truth table.
- Denoted by $M_{j}$, where $j$ is the decimal equivalent of the maxterm's corresponding binary combination ( $b_{j}$ ).
- A variable in $M_{j}$ is complemented if its value in $b_{j}$ is 1 , otherwise is uncomplemented.
- Example: Assume 3 variables ( $\mathrm{A}, \mathrm{B}, \mathrm{C}$ ), and $j=3$. Then, $\mathrm{b}_{j}=011$ and its corresponding maxterm is denoted by $M_{j}=A+B^{\prime}+C^{\prime}$


## Truth Table notation for Minterms and Maxterms

- Minterms and Maxterms are easy to denote using a truth table.
- Example:

Assume 3 variables
x,y,z
(order is fixed)

| $x$ | $y$ | $z$ | Minterm | Maxterm |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | $x^{\prime} y^{\prime} z^{\prime}=m_{0}$ | $x+y+z=M_{0}$ |
| 0 | 0 | 1 | $x^{\prime} y^{\prime} y^{\prime}=m_{1}$ | $x+y+z^{\prime}=M_{1}$ |
| 0 | 1 | 0 | $x^{\prime} y z^{\prime}=m_{2}$ | $x+y^{\prime}+z=M_{2}$ |
| 0 | 1 | 1 | $x^{\prime} y z=m_{3}$ | $x+y^{\prime}+z^{\prime}=M_{3}$ |
| 1 | 0 | 0 | $x y^{\prime} z^{\prime}=m_{4}$ | $x^{\prime}+y+z=M_{4}$ |
| 1 | 0 | 1 | $x y^{\prime} z=m_{5}$ | $x^{\prime}+y+z^{\prime}=M_{5}$ |
| 1 | 1 | 0 | $x y z^{\prime}=m_{6}$ | $x^{\prime}+y^{\prime}+z=M_{6}$ |
| 1 | 1 | 1 | $x y z=m_{7}$ | $x^{\prime}+y^{\prime}+z^{\prime}=M_{7}$ |

## Canonical Forms (Unique)

- Each term of Boolean expression contain all input variables in either true form or complemented form
- Any Boolean function $F($ ) can be expressed as a unique sum of minterms and a unique product of maxterms (under a fixed variable ordering).
- In other words, every function $F()$ has two canonical forms:
$\checkmark$ Canonical Sum-Of-Products (sum of minterms)
$\checkmark$ Canonical Product-Of-Sums (product of maxterms)
- Canonical Sum-Of-Products:

The minterms included are those $\mathrm{m}_{j}$ such that F()$=1$ in row $j$ of the truth table for $\mathrm{F}($ ).

- Canonical Product-Of-Sums:

The maxterms included are those Mj such that $\mathrm{F}(\mathrm{)}=0$ in row j of the truth table for $F()$.

## Example

- Truth table for $\mathrm{F}_{1}(\mathrm{~A}, \mathrm{~B}, \mathrm{C})$ at right
- The canonical sum-of-products form for $F_{1}$ is

$$
\begin{aligned}
& \mathrm{F}_{1}(\mathrm{~A}, \mathrm{~B}, \mathrm{C})=\mathrm{m}_{1}+\mathrm{m}_{2}+\mathrm{m}_{4}+\mathrm{m}_{6} \\
& =A^{\prime} B^{\prime} C+A^{\prime} B C^{\prime}+A B^{\prime} C^{\prime}+A B C^{\prime}
\end{aligned}
$$

- The canonical product-of-sums form for $F_{1}$ is

$$
\begin{gathered}
F_{1}(A, B, C)=M_{0} \bullet M_{3} \bullet M_{5} \bullet M_{7} \\
=(A+B+C) \bullet\left(A+B^{\prime}+C^{\prime}\right) \bullet \\
\left(A^{\prime}+B+C^{\prime}\right) \cdot\left(A^{\prime}+B^{\prime}+C^{\prime}\right) .
\end{gathered}
$$

- Observe that: $\mathrm{m}_{\mathrm{j}}=\mathrm{M}_{\mathrm{j}}{ }^{\prime}$

| $A$ | $B$ | $C$ | $F_{1}$ |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 1 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 0 | 1 |
| 1 | 1 | 1 | 0 |

## Shorthand: $\Sigma$ and $\Pi$

- $F_{1}(A, B, C)=\sum m(1,2,4,6)$, where $\sum$ indicates that this is a sum-of-products form, and $m(1,2,4,6)$ indicates that the minterms to be included are $m_{1}, m_{2}, m_{4}$, and $\mathrm{m}_{6}$.
- $F_{1}(A, B, C)=\Pi M(0,3,5,7)$, where $\Pi$ indicates that this is a product-of-sums form, and $\mathrm{M}(0,3,5,7)$ indicates that the maxterms to be included are $M_{0}, M_{3}, M_{5}$, and $\mathrm{M}_{7}$.
- Since $m_{j}=M_{j}^{\prime}$ for any $j$,
$\sum m(1,2,4,6)=\Pi M(0,3,5,7)=F_{1}(A, B, C)$


## Conversion Between Canonical Forms

- Replace $\sum$ with $\Pi$ (or vice versa) and replace those j's that appeared in the original form with those that do not.
- Example:

$$
\begin{aligned}
f_{1}(a, b, c) & =a^{\prime} b^{\prime} c+a^{\prime} b c^{\prime}+a b^{\prime} c^{\prime}+a b c^{\prime} \\
& =m_{1}+m_{2}+m_{4}+m_{6} \\
& =\sum(1,2,4,6) \\
& =\prod(0,3,5,7) \\
& =(a+b+c) \cdot\left(a+b^{\prime}+c^{\prime}\right) \cdot\left(a^{\prime}+b+c^{\prime}\right) \cdot\left(a^{\prime}+b^{\prime}+c^{\prime}\right)
\end{aligned}
$$

## Standard Forms (NOT Unique)

- If there is exists at least one term that does not contain all variables.
- Standard forms are "like" canonical forms, except that not all variables need appear in the individual product (SOP) or sum (POS) terms.
- Example

$$
f_{1}(a, b, c)=a^{\prime} b^{\prime} c+b c^{\prime}+a c^{\prime}
$$

is a standard sum-of-products form
$f_{1}(a, b, c)=(a+b+c) \cdot\left(b^{\prime}+c^{\prime}\right) \cdot\left(a^{\prime}+c^{\prime}\right)$
is a standard product-of-sums form.

## Conversion of SOP from standard to canonical form

- Expand non-canonical terms by inserting equivalent of 1 in each missing variable $x$ :

$$
\left(x+x^{\prime}\right)=1
$$

- Remove duplicate minterms
- $f_{1}(a, b, c)=a^{\prime} b^{\prime} c+b c^{\prime}+a c^{\prime}$

$$
\begin{aligned}
& =a^{\prime} b^{\prime} c+\left(a+a^{\prime}\right) b c^{\prime}+a\left(b+b^{\prime}\right) c^{\prime} \\
& =a^{\prime} b^{\prime} c+a b c^{\prime}+a^{\prime} b c^{\prime}+a b c^{\prime}+a b^{\prime} c^{\prime} \\
& =a^{\prime} b^{\prime} c+a b c^{\prime}+a^{\prime} b c+a b^{\prime} c^{\prime}
\end{aligned}
$$

## Conversion of POS from standard to canonical form

- Expand noncanonical terms by adding 0 in terms of missing variables (e.g., $x x^{\prime}=0$ ) and using the distributive law
- Remove duplicate maxterms
- $f_{1}(a, b, c)=(a+b+c) \cdot\left(b^{\prime}+c^{\prime}\right) \cdot\left(a^{\prime}+c^{\prime}\right)$

$$
=(a+b+c) \bullet\left(a a^{\prime}+b^{\prime}+c^{\prime}\right) \bullet\left(a^{\prime}+b b^{\prime}+c^{\prime}\right)
$$

$$
=(a+b+c) \bullet\left(a+b^{\prime}+c^{\prime}\right) \bullet\left(a^{\prime}+b^{\prime}+c^{\prime}\right) \bullet
$$

$$
\left(a^{\prime}+b+c^{\prime}\right) \bullet
$$

$$
=(a+b+c) \cdot\left(a+b^{\prime}+c^{\prime}\right) \cdot\left(a^{\prime}+b^{\prime}+c^{\prime}\right) \cdot\left(a^{\prime}+b+c^{\prime}\right)
$$

## Karnaugh Maps

- Karnaugh maps (K-maps) are graphical representations of boolean functions.
- One map cell corresponds to a row in the truth table.
- Also, one map cell corresponds to a minterm or a maxterm in the boolean expression
- Multiple-cell areas of the map correspond to standard terms.


## Two-Variable Map



NOTE: ordering of variables is IMPORTANT for $f\left(x_{1}, x_{2}\right), x_{1}$ is the row, $x_{2}$ is the column.
Cell 0 represents $x_{1}{ }^{\prime} x_{2}{ }^{\prime}$; Cell 1 represents $x_{1}{ }^{\prime} x_{2}$ : etc. If a minterm is present in the function, then $a 1$ is placed in the corresponding cell.

## Two-Variable Map

- Any two adjacent cells in the map differ by ONLY one variable, which appears complemented in one cell and uncomplemented in the other.
Exp: $\mathrm{m}_{0}\left(=\mathrm{x}_{1}{ }^{\prime} \mathrm{x}_{2}{ }^{\prime}\right)$ is adjacent to $\mathrm{m}_{1}\left(=\mathrm{x}_{1}{ }^{\prime} \mathrm{x}_{2}\right)$ and $\mathrm{m}_{2}$ ( $=x_{1} x_{2}{ }^{\prime}$ ) but NOT $m_{3}\left(=x_{1} x_{2}\right)$


## 2-Variable Map -- Example

- $f\left(x_{1}, x_{2}\right)=x_{1}{ }^{\prime} x_{2}{ }^{\prime}+x_{1}{ }^{\prime} x_{2}+x_{1} x_{2}{ }^{\prime}$

$$
\begin{aligned}
& =\mathrm{m}_{0}+\mathrm{m}_{1}+\mathrm{m}_{2} \\
& =\mathrm{x}_{1}^{\prime}+\mathrm{x}_{2}^{\prime}
\end{aligned}
$$

- 1s placed in K-map for specified minterms $\mathrm{m}_{0}, \mathrm{~m}_{1}, \mathrm{~m}_{2}$
- Grouping (ORing) of 1s allows simplification
- What (simpler) function is represented by each dashed rectangle?
- $x_{1}{ }^{\prime}=m_{0}+m_{1}$
- $x_{2}{ }^{\prime}=m_{0}+m_{2}$

- Note $\mathrm{m}_{0}$ covered twice


## Minimization as SOP using K-map

- Enter 1s in the K-map for each product term in the function
- Group adjacent K-map cells containing 1s to obtain a product with fewer variables. Group size must be in power of $2(2,4,8, \ldots)$
- Handle "boundary wrap" for K-maps of 3 or more variables.
- Realize that answer may not be unique


## Three-Variable Map


-Note: variable ordering is ( $x, y, z$ ); yz specifies column, $x$ specifies row.
-Each cell is adjacent to three other cells (left or right or top or bottom or edge wrap)

## Three-Variable Map (cont.)

The types of structures that are either minterms or are generated by repeated
 application of the minimization theorem on a three variable map are shown at right. Groups of 1, 2, 4, 8 are possible.

group of 2 terms



## Simplification

- Enter minterms of the Boolean function into the map, then group terms
- Example: $f(a, b, c)=a^{\prime} c+a b c+b c^{\prime}$
- Result: $f(a, b, c)=a^{\prime} c+b$



## More Examples

- $f_{1}(x, y, z)=\sum m(2,3,5,7)$

$$
-f_{1}(x, y, z)=x^{\prime} y+x z
$$



- $f_{2}(x, y, z)=\sum m(0,1,2,3,6)$

$$
\square f_{2}(x, y, z)=x^{\prime}+y z^{\prime}
$$



## Four-Variable Maps



- Top cells are adjacent to bottom cells. Left-edge cells are adjacent to right-edge cells.
- Note variable ordering (WXYZ).


## Four-variable Map Simplification

- One square represents a minterm of 4 literals.
- A rectangle of 2 adjacent squares represents a product term of 3 literals.
- A rectangle of 4 squares represents a product term of 2 literals.
- A rectangle of 8 squares represents a product term of 1 literal.
- A rectangle of 16 squares produces a function that is equal to logic 1.


## Example

- Simplify the following Boolean function (A,B,C,D) = $\sum m(0,1,2,4,5,7,8,9,10,12,13)$.
- First put the function $g()$ into the map, and then group as many 1s as possible.

| ab |
| :--- |
| 1 |
| 1 |$| 1$


$\mathbf{g}(\mathbf{A}, \mathbf{B}, \mathbf{C}, \mathbf{D})=\mathbf{c}^{\prime}+\mathbf{b}^{\prime} \mathbf{d}^{\prime}+\mathbf{a}^{\prime} \mathbf{b d}$

## Don't Care Conditions

- There may be a combination of input values which
- will never occur
- if they do occur, the output is of no concern.
- The function value for such combinations is called a don't care.
- They are denoted with $\mathbf{x}$ or - . Each $\mathbf{x}$ may be arbitrarily assigned the value 0 or 1 in an implementation.
- Don't cares can be used to further simplify a function


## hinimization using Don’t Cares

- Treat don't cares as if they are 1s to generate Pls.
- Delete Pl's that cover only don't care minterms.
- Treat the covering of remaining don't care minterms as optional in the selection process (i.e. they may be, but need not be, covered).


## Example

- Simplify the function $f(a, b, c, d)$ whose K-map is shown at the

| $a b$ |  |  |  |  |
| ---: | :--- | :--- | :--- | ---: |
| $a b$ | 00 | 01 | 11 | 10 |
| 00 | 0 | 1 | 0 | 1 |
| 00 | 1 | 1 | 0 | 1 |
| 11 | 0 | 0 | $x$ | $x$ |
| 10 | 1 | 1 | $x$ | $x$ |
|  |  |  |  |  | right.

- $f=a^{\prime} c^{\prime} d+a b^{\prime}+c d^{\prime}+a^{\prime} b c^{\prime}$
or
- $f=a^{\prime} c^{\prime} d+a b^{\prime}+c d^{\prime}+a^{\prime} b d^{\prime}$

| 0 | 1 | 0 | 1 |
| :---: | :---: | :---: | :---: |
| -1 | 1 | 0 | 1 |
| 0 | 0 | $x$ | $x$ |
| 1 | 1 | $x$ | $x$ |

## Another Example

- Simplify the function $g(a, b, c, d)$ whose K-map is shown at right.
- $\mathrm{g}=\mathrm{a}^{\prime} \mathrm{c}^{\prime}+\mathrm{ab}$ or
- $g=a^{\prime} c^{\prime}+b^{\prime} d$

$a b$| $c d$ |  |  |  |
| :--- | :--- | :--- | :--- |
| $x$ | 1 | 0 | 0 |
| 1 | $x$ | 0 | $x$ |
| 1 | $x$ | $x$ | 1 |
| 0 | $x$ | $x$ | 0 |


| $x$ | $1:$ | 0 | 0 |
| :--- | :--- | :--- | :--- |
| 1 | $x$ | 0 | $x$ |
| 1 | $x$ | $x$ | 1 |
| 0 | $x$ | $x$ | 0 |


| $x$ | 1 | 0 |
| :---: | :---: | :---: |
| $1:$ | 0 |  |
| $1:$ | 0 | 0 |
| 0 | $x$ | $\times$ |

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 Karnaugh Map- Simplification methods
- Boolean algebra(algebraic method)
- Karnaugh map(map method))
- Quine-McCluskey(tabular method)


Three- and Four-input Kanaugh maps

(a) XY

(b) $\mathrm{X}+\mathrm{Y}$

(a)

(b)

| $m_{0}$ | $m_{1}$ | $m_{3}$ | $m_{2}$ |
| :--- | :--- | :--- | :--- |
| $m_{4}$ | $m_{5}$ | $m_{7}$ | $m_{6}$ |

(a)

(b)

(b)

(a)

## Karnaugh Map (K- Map) Steps

1. Sketch a Karnaugh map grid for the given problem.in power of $2^{\mathrm{N}}$ Squares
2. Fill in the 1's and 0's from the truth table of sop or pos Boolean function
3. Circle groups of 1 's.

- Circle the largest groups of $2,4,8$, etc. first.
- Minimize the number of circles but make sure that every 1 is in a circle.

4. Write an equation using these circles.

Example) $\mathbf{F}(\mathbf{X}, \mathbf{Y}, \mathbf{Z})=\Sigma \mathbf{m}(\mathbf{2}, \mathbf{3}, 4,5)=\mathbf{X}^{\prime} \mathbf{Y}+\mathbf{X Y} \mathbf{Y}^{\prime}$


Example) $\mathbf{F}(\mathbf{X}, \mathbf{Y}, \mathbf{Z})=\Sigma \mathbf{m}(\mathbf{0}, \mathbf{2 , 4 , 6})=\mathbf{X}^{\prime} \mathbf{Z}^{\prime}+\mathbf{X Z} \mathbf{Z}^{\prime}=\mathbf{Z}^{\prime}\left(\mathbf{X}^{\prime}+\mathbf{X}\right)=\mathbf{Z}^{\prime}$


Three-Variable Map: Flat and on a Cylinder to Show Adjacent Squares

Nariable K-Map : 16 minterms : $\mathrm{m}_{0} \sim \mathrm{~m}_{15}$
Rectangle group

- 2-squares(minterms) : 3-literals product term
- 4-squares : 2-literals product term
- 8-squares : 1-literals product term
- 16 -squares : logic 1

| $m_{0}$ | $m_{1}$ | $m_{3}$ | $m_{2}$ |
| :--- | :--- | :--- | :--- |
| $m_{4}$ | $m_{5}$ | $m_{7}$ | $m_{6}$ |
| $m_{12}$ | $m_{13}$ | $m_{15}$ | $m_{14}$ |
| $m_{8}$ | $m_{9}$ | $m_{11}$ | $m_{10}$ |

(a)

(b)

Fig. 2-17 Four-Variable Map
$\mathbf{N}, \mathbf{X}, \mathbf{Y}, \mathbf{Z})=\boldsymbol{\operatorname { m }}(\mathbf{0}, \mathbf{2}, \mathbf{7}, \mathbf{8 , 9 , 1 0 , 1 1})=\mathbf{W} \mathbf{X}^{\prime}+\mathbf{X}^{\prime} \mathbf{Z}^{\prime}+\mathbf{W}^{\prime} \mathbf{X Y Z}$

| $\begin{array}{\|c\|} w_{x} \\ \hline \end{array}$ | $\begin{array}{lllll}00 & 01 & 11 & 10\end{array}$ |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| 00 | (1) | 0 | 0 | 1 |
| 01 | 0 | 0 | 1 | 0 |
| 11 | 0 | 0 | 0 | 0 |
| 10 | (1) | (1) | (1) | (1) |
| (a) |  |  |  |  |


| $w_{\text {yz }}{ }^{\text {y }}$ |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | 00 | 01 | 11 |  |
| $\frac{w x}{00}$ | 1 | 0 | 0 | 1 |
| 01 | 0 | 0 | 1 | 0 |
| 11 | 0 | 0 | 0 | 0 |
| 10 | (1) | 1 | 1 | (1) |
|  |  |  | b) |  |

Ex 4-28) Minimize the following expression $A B^{\prime} C+A^{\prime} B C+A^{\prime} B^{\prime} C+A^{\prime} B^{\prime} C^{\prime}+A B^{\prime} C^{\prime}$

Sol) $B^{\prime}+A^{\prime} C$

$E^{\circ}$ y linimize the following expression $B^{\prime} C^{\prime} D^{\prime}+A^{\prime} B C^{\prime} D^{\prime}+A B C^{\prime} D^{\prime}+A^{\prime} B^{\prime} C D+A B^{\prime} C D+A^{\prime} B^{\prime}{ }^{\prime}$ Sol) $D^{\prime}+B^{\prime} C$

- Don't Care Conditions
- it really does not matter since they will never occur(its output is either ' 0 ' or ' 1 ')
- The don't care terms can be used to advantage on the Karnaugh map

| Inputs | Output |
| :---: | :---: |
| ABCD | $\boldsymbol{r}$ |
| 0000 | 0 |
| 0001 | 0 |
| 0010 | 0 |
| 0011 | 0 |
| 0100 | 0 |
| 0101 | 0 |
| 0110 | 0 |
| 0111 | 1 |
| 1000 | 1 |
| 1001 | 1 |
| 1010 | X |
| 1011 | X |
| 1100 | X |
| 1101 | X |
| 1110 | X |
| 1111 | X |

(a) Truth table

(b) Without "don't cares" $Y=A \bar{B} \bar{C}+\bar{A} B C D$ With "don't cares" $Y=A+B C D$

ExK-Map for POS $(B+C+D)\left(A+B+C^{\prime}+D\right)\left(A^{\prime}+B+C+D^{\prime}\right)\left(A+B^{\prime}+C+D\right)\left(A^{\prime}+B^{\prime}+C+D\right)$
Sol) $(B+C+D)=\left(A^{\prime} A+B+C+D\right)=\left(A^{\prime}+B+C+D\right)(A+B+C+D)$
$(1+0+0+0)(0+0+0+0)(0+0+1+0)$
$(1+0+0+1)(0+1+0+0)(1+1+0+0)$

$$
F=(C+D)\left(A^{\prime}+B+C\right)(A+B+D)
$$

- Converting Between POS and SOP Using the K-map
Ex 4-33) $\left(A^{\prime}+B^{\prime}+C+D\right)\left(A+B^{\prime}+C+D\right)$ $\left(A+B+C+D^{\prime}\right)\left(A+B+C^{\prime}+D^{\prime}\right)\left(A^{\prime}+B+C+D^{\prime}\right)$

(a) Minimum POS: $(A+B+\bar{C})(\bar{B}+C+D)(B+C+\bar{D})$


(b) Standard SOP: $\bar{A} \bar{B} \bar{C} \bar{D}+\bar{A} B \bar{C} D+\bar{A} B C D+\bar{A} B C \bar{D}+A B C \bar{D}+A \bar{B} C \bar{D}+$ $A \bar{B} \bar{C} \bar{D}+A B \bar{C} D+A \bar{B} C D+A B C D$

(c) Minimum SOP: $A C+B C+B D+\bar{B} \bar{C} \bar{D}$
- Five Variable K-Map : $\{A, B, C, D, E\}$

- Six Variable K-Map : $\{A, B, C, D, E, F\}$


