Objectives:

- Calculate the Laplace transform of common functions using the definition and the Laplace transform tables
 Laplace-transform a circuit, including components with
- non-zero initial conditions.
- •Analyze a circuit in the s-domain
- Check your s-domain answers using the initial value theorem (IVT) and final value theorem (FVT)
 Inverse Laplace-transform the result to get the time-domain solutions; be able to identify the forced and natural response components of the time-domain solution.
 (Note this material is covered in Chapter 12 and Sections 13.1 13.3)

What types of circuits can we analyze?

- •Circuits with any number and type of DC sources and any number of resistors.
- •First-order (RL and RC) circuits with no source and with a DC source.
- •Second-order (series and parallel RLC) circuits with no source and with a DC source.

•Circuits with sinusoidal sources and any number of resistors, inductors, capacitors (and a transformer or op amp), but can generate only the <u>steady-state</u> response.

What types of circuits will Laplace methods allow us to analyze?

•Circuits with any type of source (so long as the function describing the source has a Laplace transform), resistors, inductors, capacitors, transformers, and/or op amps; the Laplace methods produce the <u>complete</u> response!

Definition of the Laplace transform:

$$\mathcal{L}{f(t)} = F(s) = \int_0^\infty f(t)e^{-st}dt$$

Note that there are limitations on the types of functions for which a Laplace transform exists, but those functions are "pathological", and not generally of interest to engineers!

Aside – formally define the "step function", which is often modeled in a circuit by a voltage source in series with a switch.



When K = 1, f(t) = u(t), which we call the unit step function

More step functions:

The step function shifted in time

The "window" function



Which of these expressions describes the function plotted here?



X A.
$$u(t - 5)$$

X B. $5u(t + 5)$

$$5.50(1 + 15)$$

15u(t-5)

Which of these expressions describes the function plotted here?







Use "window" functions to express this piecewise linear function as a single function valid for all time.





0. t < 02t, $0 \le t \le 1$ s [u(t) - u(t-1)] $f(t) = -2t+4, \quad 0 \le t \le 1s \quad [u(t-1)-u(t-3)]$ 2t-8, $0 \le t \le 1s$ [u(t-3)-u(t-4)]0. t > 4 sf(t) = 2t[u(t) - u(t-1)] + (-2t+4)[u(t-1) - u(t-3)]+(2t-8)[u(t-3)-u(t-4)]= 2tu(t) - 4(t-1)u(t-1) + 4(t-3)u(t-3) - 2(t-4)u(t-4)

The impulse function, created so that the step function's derivative is defined for all time:



Use a limiting function to define the step function and its first derivative!

The step function

The first derivative of the step function



The unit impulse function is represented symbolically as $\delta(t)$. Definition: $\delta(t) = 0$ for $t \neq 0$

$$\delta(t) = 0$$
 for $t \neq 0$
and $\int_{-\infty}^{\infty} \delta(t) dt = 1$
(Note that thearea under the $g(t)$ function is

$$\frac{1}{2\varepsilon}(\varepsilon + \varepsilon)$$
, which approaches 1 as $\varepsilon \to 0$)

Note also that any limiting function with the following characteristics can be used to generate the unit impulse function:

•Height
$$\rightarrow \infty$$
 as $\epsilon \rightarrow 0$

- •Width \rightarrow 0 as $\epsilon \rightarrow$ 0
- •Area is constant for all values of $\boldsymbol{\epsilon}$



The sifting property is an important property of the impulse function:

$$\int_{-\infty}^{\infty} f(t)\delta(t-a)dt = f(a)$$

Evaluate the following integral, using the sifting property of the impulse function.

$$\int_{-10}^{10} (6t^2 + 3)\delta(t - 2)dt$$



$$6(2)^2 + 3 = 27$$

Use the definition of Laplace transform to calculate the Laplace transforms of some functions of interest:

$$\mathcal{L}\{\delta(t)\} = \int_0^\infty \delta(t)e^{-st}dt = \int_0^\infty \delta(t-0)e^{-st}dt = e^{-s(0)} = 1$$
$$\mathcal{L}\{u(t)\} = \int_0^\infty u(t)e^{-st}dt = \int_0^\infty 1e^{-st}dt = \frac{1}{-s}e^{-st}\Big|_0^\infty = 0 - \frac{1}{-s} = \frac{1}{s}$$

$$\mathcal{L}\{e^{-at}\} = \int_0^\infty e^{-at} \bar{e}^{st} dt = \int_0^\infty e^{-(s+a)t} dt = \frac{1}{-(s+a)} e^{-(s+a)t} \bigg|_0^\infty = 0 - \frac{\Box 1}{-(s+a)} - \frac{\Box 1}{-(s+a)} (s+a)$$

$$\mathcal{L}\{\sin\omega t\} = \int_0^\infty \left[\frac{e^{j\omega t} - e^{-j\omega t}}{2j}\right] e^{-st} dt = \frac{1}{j^2} \int_0^\infty \left[e^{-(s-j\omega)t} - e^{-(s+j\omega)t}\right] dt$$
$$= \frac{1}{j^2} \left[\frac{e^{-(s-j\omega)t}}{-(s-j\omega)}\right]_0^\infty - \frac{1}{j^2} \left[\frac{e^{-(s+j\omega)t}}{-(s+j\omega)}\right]_0^\infty = \frac{1}{j^2} \left[\frac{1}{(s-j\omega)} - \frac{1}{(s+j\omega)}\right] = \frac{\omega}{s^2 + \omega^2}$$

Look at the Functional Transforms table. Based on the pattern that exists relating the step and ramp transforms, and the exponential and damped-ramp transforms, what do you predict the Laplace transform of t² is?

$$\begin{array}{c} \mathbf{X} \quad \mathbf{A}. \quad 1/(\mathbf{s} + \mathbf{a}) \\ \mathbf{X} \quad \mathbf{B}. \quad \mathbf{s} \\ \mathbf{\sqrt{c}}. \quad 1/\mathbf{s}^3 \end{array}$$

Using the definition of the Laplace transform, determine the effect of various operations on time-domain functions when the result is Laplace-transformed. These are collected in the Operational Transform table.

$$\begin{aligned} \mathcal{L}\{K_{1}f_{1}(t) + K_{2}f_{2}(t) - K_{3}f_{3}(t)\} &= \int_{0}^{\infty} [K_{1}f_{1}(t)e^{-st} + K_{2}f_{2}(t)e^{-st} - K_{3}f_{3}(t)e^{-st}]dt \\ &= \int_{0}^{\infty} K_{1}f_{1}(t)e^{-st}dt + \int_{0}^{\infty} K_{2}f_{2}(t)e^{-st}dt - \int_{0}^{\infty} K_{3}f_{3}(t)e^{-st}dt \\ &= K_{1}\int_{0}^{\infty} f_{1}(t)e^{-st}dt + K_{2}\int_{0}^{\infty} f_{2}(t)e^{-st}dt - K_{3}\int_{0}^{\infty} f_{3}(t)e^{-st}dt \\ &= K_{1}F_{1}(s) + K_{2}F_{2}(s) - K_{2}F_{2}(s) \\ \\ \mathcal{L}\left\{\frac{df(t)}{dt}\right\} = e^{-st}f(t)\Big|_{0}^{\infty} - \int_{0}^{\infty} f(t)[-se^{-st}]dt \qquad \text{(integration by parts!)} \\ &= -f(0) + s\int_{0}^{\infty} f(t)e^{-st}dt = sF(s) - f(0) \end{aligned}$$

Now lets use the operational transform table to find the correct value of the Laplace transform of t², given that

$$\mathcal{L}{t} = \frac{1}{s^2}$$

X A.
$$1/s^3$$

B. $2/s^3$
X C. $-2/s^3$

Example – Find the Laplace transform of $t^2e^{-\alpha t}$.

Use the operational transform: $\mathcal{L}\left\{t^{n}f(t)\right\} = (-1)^{n} \frac{d^{n}F(s)}{ds^{n}}$ Use the functional transform: $\mathcal{L}\left\{t^{-at}\right\} = \frac{1}{(s+a)}$

$$\mathcal{L}\left\{te^{-at}\right\} = (-1)^2 \frac{d^2}{ds^2} \left[\frac{\Box 1}{s+a}\right] = \frac{d}{ds} \left[\frac{\Box - 1}{(s+a)^2}\right] = \frac{\Box 2}{(s+a)^3}$$

Alternatively, Use the operational transform: $\mathcal{L}\left\{e^{-at}f(t)\right\} = F(s+a)$

Use the functional transform:

$$\mathcal{L}\left\{e^{-at}\right\} = \frac{2}{(s+a)^3}$$

How can we use the Laplace transform to solve circuit problems?

- •Option 1:
 - •Write the set of differential equations in the time domain that describe the relationship between voltage and current for the circuit.
 - •Use KVL, KCL, and the laws governing voltage and current for resistors, inductors (and coupled coils) and capacitors.
 - •Laplace transform the equations to eliminate the integrals and derivatives, and solve these equations for V(s) and I(s).
 - •Inverse-Laplace transform to get v(t) and i(t).

How can we use the Laplace transform to solve circuit problems?

•Option 2:

•Laplace transform the circuit (following the process we used in the phasor transform) and use DC circuit analysis to find V(s) and I(s).

•Inverse-Laplace transform to get v(t) and i(t).







Find the value of the complex impedance and the series-connected voltage source, representing the Laplace transform of a capacitor.

X A.
$$sC, V_0/s$$

B. $1/sC, V_0/s$
C. $1/sC, -V_0/s$



 $I(s) = sCV(s) - CV_0$

Recipe for Laplace transform circuit analysis:

- 1. Redraw the circuit (nothing about the Laplace transform changes the types of elements or their interconnections).
- 2. Any voltages or currents with values given are Laplacetransformed using the functional and operational tables.
- Any voltages or currents represented symbolically, using i(t) and v(t), are replaced with the symbols I(s) and V(s).
- 4. All component values are replaced with the corresponding complex impedance, Z(s).
- 5. Use DC circuit analysis techniques to write the s-domain equations and solve them.
- 6. Inverse-Laplace transform s-domain solutions to get timedomain solutions.

Example:

There is no initial energy stored in this circuit. Find $i_1(t)$ and $i_2(t)$ for t > 0.



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Recipe for Laplace transform circuit analysis:

- 1. Redraw the circuit (nothing about the Laplace transform changes the types of elements or their interconnections).
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- 4. All component values are replaced with the corresponding complex impedance, Z(s).
- 5. Use DC circuit analysis techniques to write the s-domain equations and solve them.
- 6. Inverse-Laplace transform s-domain solutions to get timedomain solutions.

Finding the inverse Laplace transform:

$$f(t) = \frac{1}{j2\pi} \int_{c-j\infty}^{c+j\infty} F(s) e^{st} ds \qquad t > 0$$

This is a contour integral in the complex plane, where the complex number c must be chosen such that the path of integration is in the convergence area along a line parallel to the imaginary axis at distance c from it, where c must be larger than the real parts of all singular values of F(s)!

There must be a better way ...

Inverse Laplace transform using partial fraction expansion: •Every s-domain quantity, V(s) and I(s), will be in the form $\frac{N(s)}{D(s)}$

where N(s) is the numerator polynomial in s, and has real coefficients, and D(s) is the denominator polynomial in s, and also has real coefficients, and O(N(s)) < O(D(s))

 $O\{N(s)\} < O\{D(s)\}$

•Since D(s) has real coefficients, it can always be factored, where the factors can be in the following forms:

- ✓ Real and distinct
- \checkmark Real and repeated
- ✓ Complex conjugates and distinct
- ✓ Complex conjugates and repeated

Inverse Laplace transform using partial fraction expansion:

- •The roots of D(s) (the values of s that make D(s) = 0) are called **poles**.
- •The roots of N(s) (the values of s that make N(s) = 0) are called zeros.

Back to the example:

$$I_1(s) = \frac{40s + 360}{s^3 + 14s^2 + 24s} = \frac{40(s+9)}{s(s+2)(s+12)}$$
$$I_2(s) = \frac{168}{s^3 + 14s^2 + 24s} = \frac{168}{s(s+2)(s+12)}$$

Find the zeros of $I_1(s)$.

$$I_1(s) = \frac{40(s+9)}{s(s+2)(s+12)}$$



X c. There aren't any zeros

Find the poles of
$$I_1(s)$$
.

$$I_1(s) = \frac{40(s+9)}{s(s+2)(s+12)}$$

Example:

There is no initial energy stored in this circuit. Find $i_1(t)$ and $i_2(t)$ for t > 0.

$$I_1(s) = \frac{40s + 360}{s(s+2)(s+12)}$$
$$= \frac{K_1}{s} + \frac{K_2}{s+2} + \frac{K_3}{s+12}$$



$$\therefore \qquad I_1(s) = \frac{15}{s} + \frac{-14}{s+2} + \frac{-1}{s+12}$$

Example:

There is no initial energy stored in this circuit. Find $i_1(t)$ and $i_2(t)$ for t > 0.



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$$i_1(t) = \mathcal{L}^{-1}\left\{\frac{15}{s} + \frac{-14}{s+2} + \frac{-1}{s+12}\right\}$$
$$= [15 - 14e^{-2t} - e^{-12t}]u(t) A$$

Theforced response is 15u(t) A;

The natural response is $[-14e^{-2t} - e^{-12t}]u(t)$ A.
Example:

There is no initial energy stored in this circuit. Find $i_1(t)$ and $i_2(t)$ for t > 0.

$$I_2(s) = \frac{168}{s(s+2)(s+12)}$$
$$= \frac{K_1}{s} + \frac{K_2}{s+2} + \frac{K_3}{s+12}$$



$$K_{1} = \frac{168}{(s+2)(s+12)} \bigg|_{s=0} = 7; \qquad K_{2} = \frac{168}{s(s+12)} \bigg|_{s=-2} = -8.4; \qquad K_{3} = \frac{168}{s(s+2)} \bigg|_{s=-12} = 1.4$$

$$\therefore \qquad I_{2}(s) = \frac{7}{s} + \frac{-8.4}{s+2} + \frac{1.4}{s+12}$$

Example:

There is no initial energy stored in this circuit. Find $i_1(t)$ and $i_2(t)$ for t > 0.



 $i_{2}(t) = \mathcal{L}^{-1} \left\{ \frac{7}{s} + \frac{-8.4}{s+2} + \frac{1.4}{s+12} \right\}$ $= [7 - 8.4e^{-2t} + 1.4e^{-12t}]u(t) A$

Theforced response is 7u(t) A;

The natural response is $[-8.4e^{-2t} - 1.4e^{-12t}]u(t)$ A.

Example:

There is no initial energy stored in this circuit. Find $i_1(t)$ and $i_2(t)$ for t > 0.

• 1 .)



$$i_{1}(t) = (15 - 14e^{-2t} - e^{-12t})u(t)A$$
$$i_{2}(t) = (7 - 8.4e^{-2t} + 1.4e^{-12t})u(t)A$$

_?t

Check the answers at t = 0 and $t = \infty$ to make sure the circuit and the equations match!

Example:

There is no initial energy stored in this circuit. Find $i_1(t)$ and $i_2(t)$ for t > 0.



$$i_{1}(t) = (15 - 14e^{-2t} - e^{-12t})u(t)A$$
$$i_{2}(t) = (7 - 8.4e^{-2t} + 1.4e^{-12t})u(t)A$$

At t = 0, the circuit has no initial stored energy, so $i_1(0) = 0$ and $i_2(0) = 0$. Now check the equations:

$$i_1(0) = (15 - 14 - 1)(1) = 0$$

 $i_2(0) = (7 - 8.4 + 1.4)(1) = 0$

As $t \to \infty$, the inductors behave like



- Inductors
- **X** B. Open circuits
 - Short circuits



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Example:

There is no initial energy stored in this circuit. Find $i_1(t)$ and $i_2(t)$ for t > 0.



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 $i_{1}(t) = (15 - 14e^{-2t} - e^{-12t})u(t)A \implies i_{1}(\infty) = 15 - 0 - 0 = 15A$ $i_{2}(t) = (7 - 8.4e^{-2t} + 1.4e^{-12t})u(t)A \implies i_{2}(\infty) = 7 - 0 - 0 = 7A$

Draw the circuit for $t = \infty$ and check these solutions.

$$42 \parallel 48 = 22.4 \Omega$$

$$336 \text{ V} \underbrace{+}_{i_1(\infty)}^{i_1(\infty)} \underbrace{i_2(\infty)}_{42 \Omega}^{i_2(\infty)} \underbrace{+}_{48 \Omega}^{i_1(\infty)} = \frac{336}{22.4} = 15 \text{ A(check!)}$$
$$i_2(\infty) = \frac{22.4}{48} (15) = 7 \text{ A(check!)}$$

We can also check the initial and final values in the s-domain, before we begin the process of inverse-Laplace transforming our s-domain solutions. To do this, use the **Initial Value Theorem (IVT)** and the **Final Value Theorem (FVT)**.

•The initial value theorem:

 $\lim_{t\to 0^+} f(t) = \lim_{s\to\infty} sF(s)$

This theorem is valid if and only if f(t) has no impulse functions.

•The final value theorem:

$$\lim_{t \to \infty} f(t) = \lim_{s \to 0} sF(s)$$

This theorem is valid if and only if all but one of the poles of F(s) are in the left-half of the complex plane, and the one that is not can only be at the origin.

Example:

There is no initial energy stored in this circuit. Find $i_1(t)$ and $i_2(t)$ for t > 0.

$$I_1(s) = \frac{40s + 360}{s^3 + 14s^2 + 24s}$$
$$I_2(s) = \frac{168}{s^3 + 14s^2 + 24s}$$



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Check your answers using the IVT and the FVT.

IVT:

From the circuit, $i_1(0) = 0$ and $i_2(0) = 0$.



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$$I_{1}(s) = \frac{40s + 360}{s^{3} + 14s^{2} + 24s}$$
$$\lim_{t \to 0} i_{1}(t) = \lim_{s \to \infty} sI_{1}(s)$$
$$= \lim_{s \to \infty} \frac{40s^{2} + 360s}{s^{3} + 14s^{2} + 24s}$$
$$= \lim_{1/s \to 0} \frac{(40/s) + (360/s^{2})}{1 + (14/s) + (24/s^{2})}$$
$$= 0 \text{A(check!)}$$

$$I_{2}(s) = \frac{168}{s^{3} + 14s^{2} + 24s}$$
$$\lim_{t \to \infty} i_{1}(t) = \lim_{s \to \infty} sI_{1}(s)$$
$$= \lim_{s \to \infty} \frac{168s}{s^{3} + 14s^{2} + 24s}$$
$$= \lim_{1/s \to 0} \frac{(168/s^{2})}{1 + (14/s) + (24/s^{2})}$$
$$= 0 \text{A(check!)}$$

FVT:

From the circuit, $i_1(\infty) = 15$ A and $i_2(\infty) = 7$ A.



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$$I_{1}(s) = \frac{40s + 360}{s^{3} + 14s^{2} + 24s}$$
$$\lim_{t \to \infty} i_{1}(t) = \lim_{s \to 0} sI_{1}(s)$$
$$= \lim_{s \to 0} \frac{40s^{2} + 360s}{s^{3} + 14s^{2} + 24s}$$
$$= \lim_{s \to 0} \frac{40s + 360}{s^{2} + 14s + 24}$$
$$= \frac{360}{24} = 15 \text{ A(check!)}$$

$$I_{2}(s) = \frac{168}{s^{3} + 14s^{2} + 24s}$$
$$\lim_{t \to \infty} i_{1}(t) = \lim_{s \to 0} sI_{1}(s)$$
$$= \lim_{s \to 0} \frac{168s}{s^{3} + 14s^{2} + 24s}$$
$$= \lim_{s \to 0} \frac{168}{s^{2} + 14s + 24}$$
$$= \frac{168}{24} = 7 \text{ A(check!)}$$

Recipe for Laplace transform circuit analysis:

- 1. Redraw the circuit (nothing about the Laplace transform changes the types of elements or their interconnections).
- 2. Any voltages or currents with values given are Laplacetransformed using the functional and operational tables.
- 3. Any voltages or currents represented symbolically, using i(t) and v(t), are replaced with the symbols I(s) and V(s).
- 4. All component values are replaced with the corresponding complex impedance, Z(s).
- 5. Use DC circuit analysis techniques to write the s-domain equations and solve them. Check your solutions with IVT and FVT.
- 6. Inverse-Laplace transform s-domain solutions to get timedomain solutions. Check your solutions at t = 0 and $t = \infty$.

Example: Find $v_0(t)$ for t > 0.



Begin by finding the initial conditions for this circuit.

$$V_o = 0 \text{ V}$$

 $I_o = \frac{70}{350} = 0.2 \text{ A}$

Give the basic interconnections of this circuit, should we use a voltage source or a current source to represent the initial condition for the inductor?



- Voltage source
- **X**_{B.} Current source
- X c. Doesn't matter



Example: Find $v_0(t)$ for t > 0.

 $\frac{1}{s512 n}\Omega$

70 V

Laplace transform the circuit and solve for $V_0(s)$.

350 Ω

I(s)

V(s)





Use the IVT and FVT to check $V_0(s)$.



Example: Find $v_0(t)$ for t > 0.

IVT $V_0(s) = \frac{70s - 268,125}{s^2 + 1750s + 9,765,625}$ $\lim_{t \to 0} v_o(t) = \lim_{s \to \infty} sV_o(s)$ $= \lim_{s \to \infty} \frac{70s^2 - 268,125s}{s^2 + 1750s + 9,765,625}$ $= \lim_{1/s \to 0} \frac{70 - 268,125/s}{1 + 1750/s + 9,765,625/s^2}$

$$= \frac{70}{1} = 70 \text{ V(check!)}$$



$$V_0(s) = \frac{70s - 268,125}{s^2 + 1750s + 9,765,625}$$
$$\lim_{t \to \infty} v_o(t) = \lim_{s \to 0} sV_o(s)$$
$$= \lim_{s \to 0} \frac{70s^2 - 268,125s}{s^2 + 1750s + 9,765,625}$$
$$= \lim_{s \to 0} \frac{0}{9,765,625}$$
$$= 0 \text{ V(check!)}$$



$$K_{1} = \frac{70s - 268,125}{(s + 875 + j3000)} \bigg|_{s = -875 + j3000} = \frac{70(-875 + j3000) - 268,125}{[(-875 + j3000) + 875 + j3000]} = 65.1 \angle 57.48^{\circ}$$

$$K_{2} = \frac{70s - 268,125}{(s + 875 - j3000)} \bigg|_{s = -875 - j3000} = \frac{70(-875 - j3000) - 268,125}{[(-875 - j3000) + 875 + -j3000]} = 65.1 \angle -57.48^{\circ}$$

When two partial fraction denominators are complex conjugates, their numerators are



🗶 🗛 Equal

X B. Unrelated

Complex conjugates

Aside – look at the inverse Laplace transform of partial fractions that are complex conjugates.

$$F(s) = \frac{10s}{s^2 + 2s + 5} = \frac{\Box_1 K}{s + 1 - j2} \frac{K^*}{s + 1 - j2} \frac{K^*}{s + 1 + j2}$$

$$K_1 = \frac{10s}{s + 1 + j2} \bigg|_{s=-1+j2} = \frac{10(-1 + j2)}{-1 + j2 + 1 + j2} = 5.59 \angle 26.57^\circ$$

$$\therefore F(s) = \frac{5.59 \angle 26.57^\circ}{s + 1 - j2} + \frac{5.59 \angle -26.57^\circ}{s + 1 + j2}$$

$$\Rightarrow f(t) = 5.59e^{j26.57^\circ}e^{-(1 - j2)t} + 5.59e^{-j26.57^\circ}e^{-(1 + j2)t}$$

$$= 5.59e^{-t}e^{j(2t + 26.57^\circ)} + 5.59e^{-t}e^{-j(2t + 26.57^\circ)}$$

$$= 5.59e^{-t}[\cos(2t + 26.57^\circ) + j\sin(2t + 26.57^\circ)]$$

$$+ 5.59e^{-t}[\cos(2t + 26.57^\circ) - j\sin(2t + 26.57^\circ)]$$

$$= 2(5.59)e^{-t}\cos(2t + 26.57^\circ)$$

The parts of the time-domain expression come from a single partial fraction term:

$$F(s) = \frac{5.59 \angle 26.57^{\circ}}{s+1-j2} + \frac{5.59 \angle -26.57^{\circ}}{s+1+j2}$$
$$f(t) = 2(5.59)e^{-t}\cos(2t+26.57^{\circ})$$

Important – you must use the numerator of the partial fraction whose denominator has the negative imaginary

The general Laplace transform (from the table below the "Functional Transforms" table)

$$F(s) = \frac{|K| \angle \theta}{s + a - jb} + \frac{|K| \angle -\theta}{s + a - jb}$$
$$\mathcal{L}^{-1} \{F(s)\} = f(t) = 2 |K| e^{-at} \cos(bt + \theta)$$

$$V_0(s) = \frac{65.1\angle 57.48^\circ}{(s+875-j3000)} + \frac{65.1\angle -57.48^\circ}{(s+875+j3000)}$$

The partial fraction expansion for $V_0(s)$ is shown above. When we inverse-Laplace transform, which partial fraction term should we use?



- The first term
- **X** B. The second term
- X c. It doesn't matter

$$V_0(s) = \frac{65.1\angle 57.48^\circ}{(s+875-j3000)} + \frac{65.1\angle -57.48^\circ}{(s+875+j3000)}$$

The time-domain function for $v_o(t)$ will include a cosine at what frequency?



X A. 875 rad/s

X в. 130.2 rad/s

✓ c. 3000 rad/s



Inverse Laplace transform:

 $v_0(t) = 2(65.1)e^{-875t}\cos(3000t + 57.48^\circ) = 130.2e^{-875t}\cos(3000t + 57.48^\circ)$ V

Check at t = 0 and t $\rightarrow \infty$: $v_0(0) = 130.2(1)\cos(57.48^\circ) = 70 \text{ V}$ $v_0(\infty) = 130.2(0)\cos(\ldots) = 0 \text{ V}$

This example is a series RLC circuit. Its response form, repeated below, is characterized as:

$$v_0(t) = 130.2e^{-875t}\cos(3000t + 57.48^\circ)$$
 V



- 🗸 🗛 Underdamped
- **X** B. Overdamped
- X c. Critically damped

Example: There is no initial energy stored in this circuit. Find v_o if $i_a = 5u(t) mA$.



Laplace transform the circuit:





$$V_o = \frac{1.4s + 20,000}{s^2 + 20,000s + 10^8}$$

This s-domain expression has _____ zeros and _____ poles.





Warning – this one's tricky!

Just after t = 0, there is no initial stored energy in the circuit. Therefore, the capacitor behaves like a _____and the inductor behaves like a ____.





- Open circuit/short circuit
- **X** B. Open circuit/open circuit
- X c. Short circuit/short circuit
 - Short circuit/open circuit



Example: Partial fraction expansion:

$$V_0(s) = \frac{1.4s + 20,000}{s^2 + 20,000s + 10^8} = \frac{1.4s + 20,000}{(s + 10,000)^2}$$
$$= \frac{K_1}{(s + 10,000)^2} + \frac{K_2}{(s + 10,000)}$$

$$V_0(s) = \frac{K_1}{(s+10,000)^2} + \frac{K_2}{(s+10,000)}$$

In the partial fraction expansion given here, K_1 and K_2 are



- Both real numbers
- Complex conjugates
- Need more information

Aside – find the partial fraction expansion when there are repeated real roots.

$$F(s) = \frac{4s^2 + 7s + 1}{s(s+1)^2} = \frac{K_1}{s} + \frac{K_2}{(s+1)^2} + \frac{K_3}{(s+1)^2} + \frac{K_3}{(s+1)^2}$$

$$K_2 = \frac{4s^2 + 7s + 1}{s} \bigg|_{s=-1} = \frac{4 - 7 + 1}{-1} = 2$$

$$K_3 = \frac{4s^2 + 7s + 1}{s(s+1)}\Big|_{s=-1} = \frac{4 - 7 + 1}{(-1)(0)} =$$
undefined!

Aside – find the partial fraction expansion when there are repeated real roots. How do we find the coefficient of the term with just one copy of the repeated root?



Aside – find the partial fraction expansion when there are repeated real roots.

$$F(s) = \frac{4s^2 + 7s + 1}{s(s+1)^2} = \frac{K_1}{s} + \frac{R_2}{(s+1)^2} + \frac{R_3}{s+1}$$

$$K_1 = \frac{4s^2 + 7s + 1}{(s+1)^2} \bigg|_{s=0} = \frac{4(0)^2 + 7(0) + 1}{(0+1)} = 1$$

$$K_2 = \frac{4s^2 + 7s + 1}{s} \bigg|_{s=-1} = \frac{4(-1)^2 + 7(-1) + 1}{(-1)} = 2$$

$$K_3 = \frac{d}{ds} \bigg[\frac{4s^2 + 7s + 1}{s} \bigg]_{s=-1} = \bigg[\frac{8s + 7}{s} - \frac{4s^2 + 7s + 1}{s^2} \bigg]_{s=-1}$$

$$= \frac{8(-1) + 7}{(-1)} - \frac{4(-1)^2 + 7(-1) + 1}{(-1)^2} = 3$$
Back to the example; find the partial fraction expansion:

$$V_0(s) = \frac{1.4s + 20,000}{(s+10,000)^2} = \frac{K_1}{(s+10,000)^2} + \frac{K_2}{(s+10,000)}$$
$$K_1 = 1.4s + 20,000 \Big|_{s=-10,000} = 6000$$

$$K_2 = \frac{d}{ds} \left[1.4s + 20,000 \right] \Big|_{s=-10,000} = 1.4$$

Example: Find $v_0(t)$ for t > 0.

Inverse Laplace transform the result in the s-domain to get the time-domain result:



 $v_0(t) = \left[6000te^{-10,000t} + 1.4e^{-10,000t} \right] u(t) \text{ V (see the Laplace tables)}$ $v_0(0) = 1.4 \text{ V (check!)}$ $v_0(\infty) = 0 \text{ V (check!)}$

$$v_o(t) = [6000te^{-10,000t} + 1.4e^{-10,000t}]u(t)$$
 V

We have seen this response form in our analysis of second-order RLC circuits; it is called:



- X A. Overdamped
- **X** B. Underdamped
- Critically damped

Example:

There is no initial energy stored in this circuit. Find i(t) if $v(t) = e^{-0.6t}sin0.8t$ V.

Laplace transform the circuit:

$$\begin{array}{c|c} & 0.96 \Omega & 1.25 \text{ F} \\ \hline t = 0 & \hline \\ + v(t) & \hline \\ i(t) \end{array} \right\} 0.8 \text{ H}$$

$$\mathcal{L}\left[e^{-0.6t}\sin 0.8t\right] = \frac{0.8}{(s+0.6)^2 + 0.8^2}$$
$$= \frac{0.8}{s^2 + 1.2s + 1}$$



Example: Find I(s):



$$\left(0.96 + \frac{0.8}{s} + 0.8s \right) I(s) = \frac{0.8}{s^2 + 1.2s + 1} \therefore \left(\frac{0.8s^2 + 0.96s + 0.8}{s} \right) I(s) = \frac{0.8}{s^2 + 1.2s + 1} \Rightarrow I(s) = \frac{s}{s^2 + 1.2s + 1}$$

 $\overline{(s^2+1.2s+1)}$

Example: Check your s-domain answer:

IVT $I(s) = \frac{s}{(s^2 + 1.2s + 1)^2}$ $\lim i(t) = \lim sI(s)$ $t \rightarrow 0$ $s \rightarrow \infty$ $= \lim_{s \to \infty} \frac{s^2}{(s^2 + 1.2s + 1)^2}$ $=\lim_{1/s\to 0}\frac{1/s^2}{(1+1.2/s+1/s^2)^2}=0$







Example:

There is no initial energy stored in this circuit. Find i(t) if $v(t) = e^{-0.6t}sin0.8t$ V.



Inverse Laplace transform the result in the s-domain to get the time-domain result:

$$I(s) = \frac{0.39 \angle -53.13^{\circ}}{(s+0.6-j0.8)^2} + \frac{\Box 0.29 \angle 90^{\circ}}{(s+0.6-j0.8)} + \dots$$

$$i(t) = 2(0.39)te^{-0.6t}\cos(0.8t - 53.13^{\circ}) + 2(0.29)e^{-0.6t}\cos(0.8t + 90^{\circ})$$

$$= \left[0.78te^{-0.6t}\cos(0.8t - 53.13^{\circ}) + 0.58e^{-0.6t}\cos(0.8t + 90^{\circ}) \right] u(t) \text{ A}$$

Which term of the solution represents the forced response?

Example:

There is no initial energy stored in this circuit. Find i(t) if $v(t) = e^{-0.6t}sin0.8t$ V.



 $i(t) = [0.78te^{-0.6t}\cos(0.8t - 53.13^\circ) + 0.58e^{-0.6t}\cos(0.8t + 90^\circ)]u(t)A$

🗶 🗛 🛛 First term

B. Second term

C. Neither

Recipe for Laplace transform circuit analysis:

- 1. Redraw the circuit note that you need to find the initial conditions and decide how to represent them in the circuit.
- 2. Any voltages or currents with values given are Laplace-transformed using the functional and operational tables.
- 3. Any voltages or currents represented symbolically, using i(t) and v(t), are replaced with the symbols I(s) and V(s).
- 4. All component values are replaced with the corresponding complex impedance, Z(s), and the appropriate source representing initial conditions.
- 5. Use DC circuit analysis techniques to write the s-domain equations and solve them. Check your solutions with IVT and FVT.
- 6. Inverse-Laplace transform s-domain solutions (using the partial fraction expansion technique and the Laplace tables) to get time-domain solutions. Check your solutions at t = 0 and $t = \infty$.

Aside – How do you inverse Laplace transform F(s) if it is an improper rational function? (Note – this won't happen in linear circuits, but can happen in other systems modeled with differential equations!) Example:

$$\mathcal{L}^{-1}\left\{\frac{s^2+6s+7}{(s+1)(s+2)}\right\}$$

(Note: $O{D(s)} > O{N(s)}$ does not hold!)

See next slide!

$$\mathcal{L}^{1}\left\{\frac{s^{2}+6s+7}{(s+1)(s+2)}\right\}$$

$$s^{2}+3s+2\overline{)s^{2}+6s+7}$$

$$\underline{-s^{2}+3s+2}$$

$$3s+5$$

(Note: $O{D(s)} > O{N(s)}$ does not hold!)

